# Calibration (I)

20th NRAO Synthesis Imaging Summer School ian.heywood@physics.ox.ac.uk



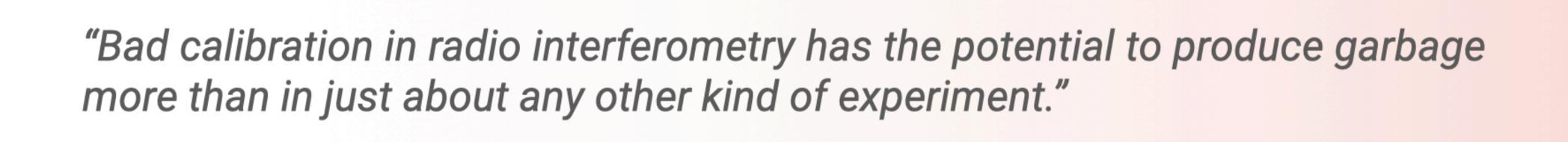












"Calibration is a dark art!"

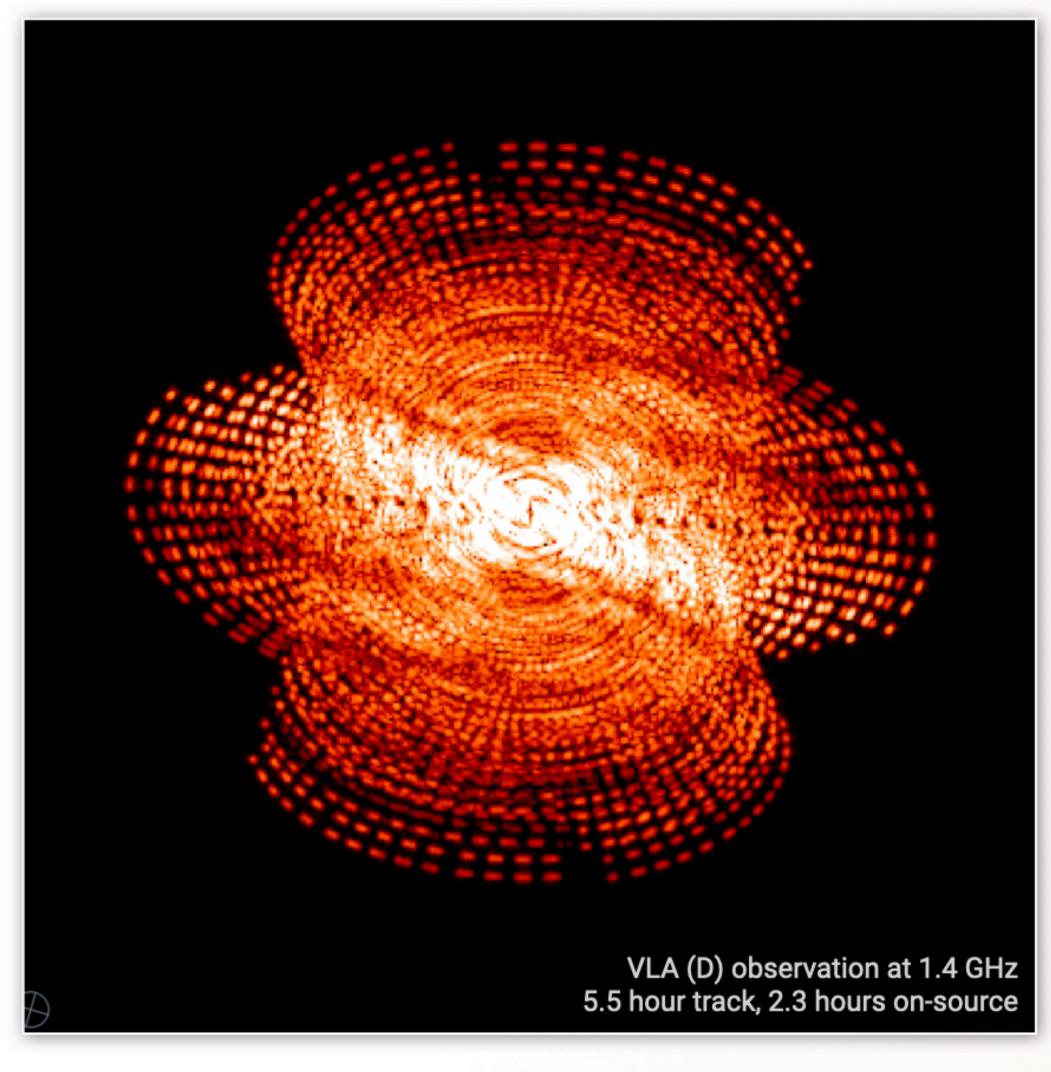
"No, it isn't."

Recall the van Cittert-Zernike theorem...

Each antenna pair in an array samples a Fourier component of the radio sky brightness distribution







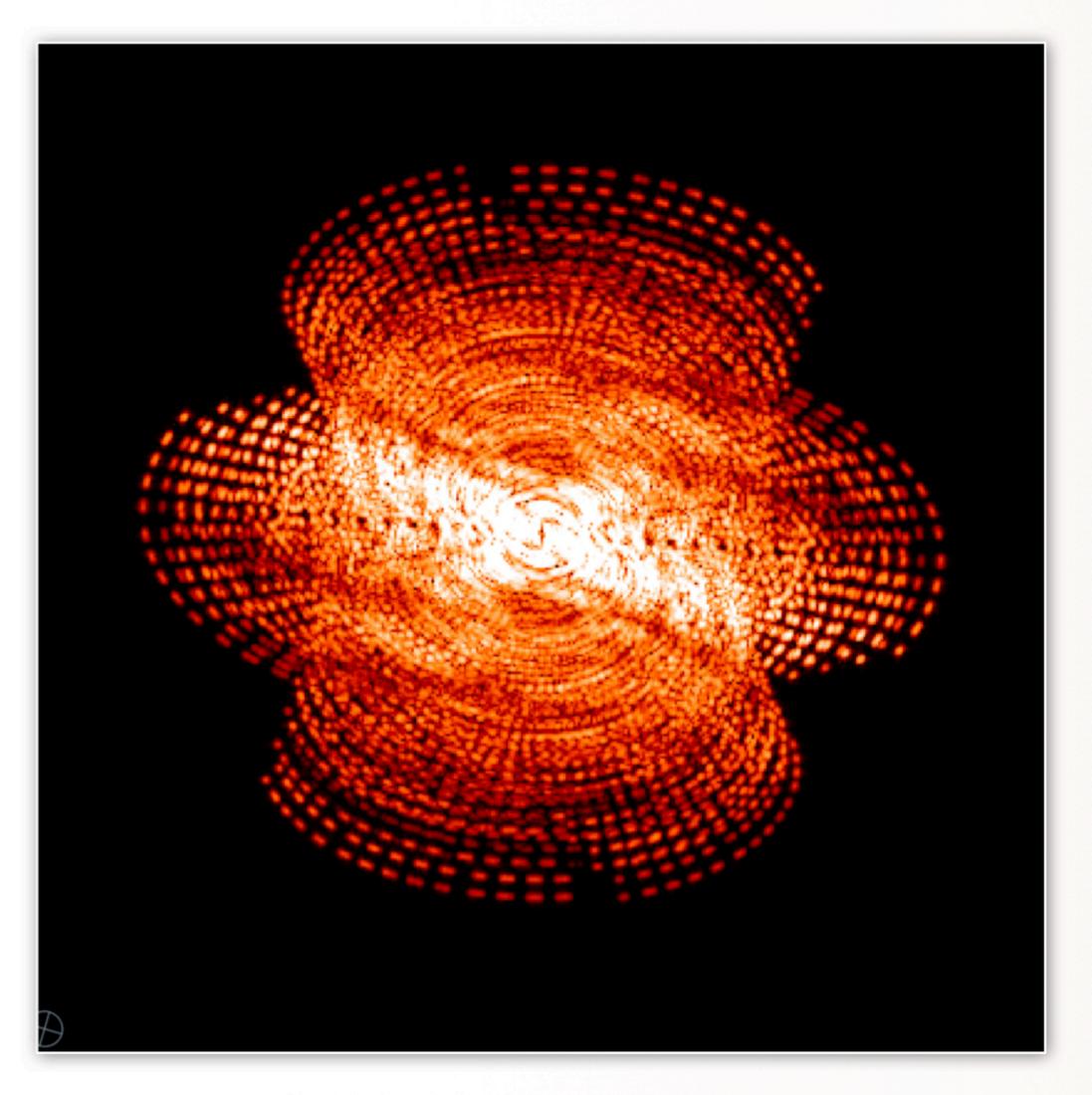
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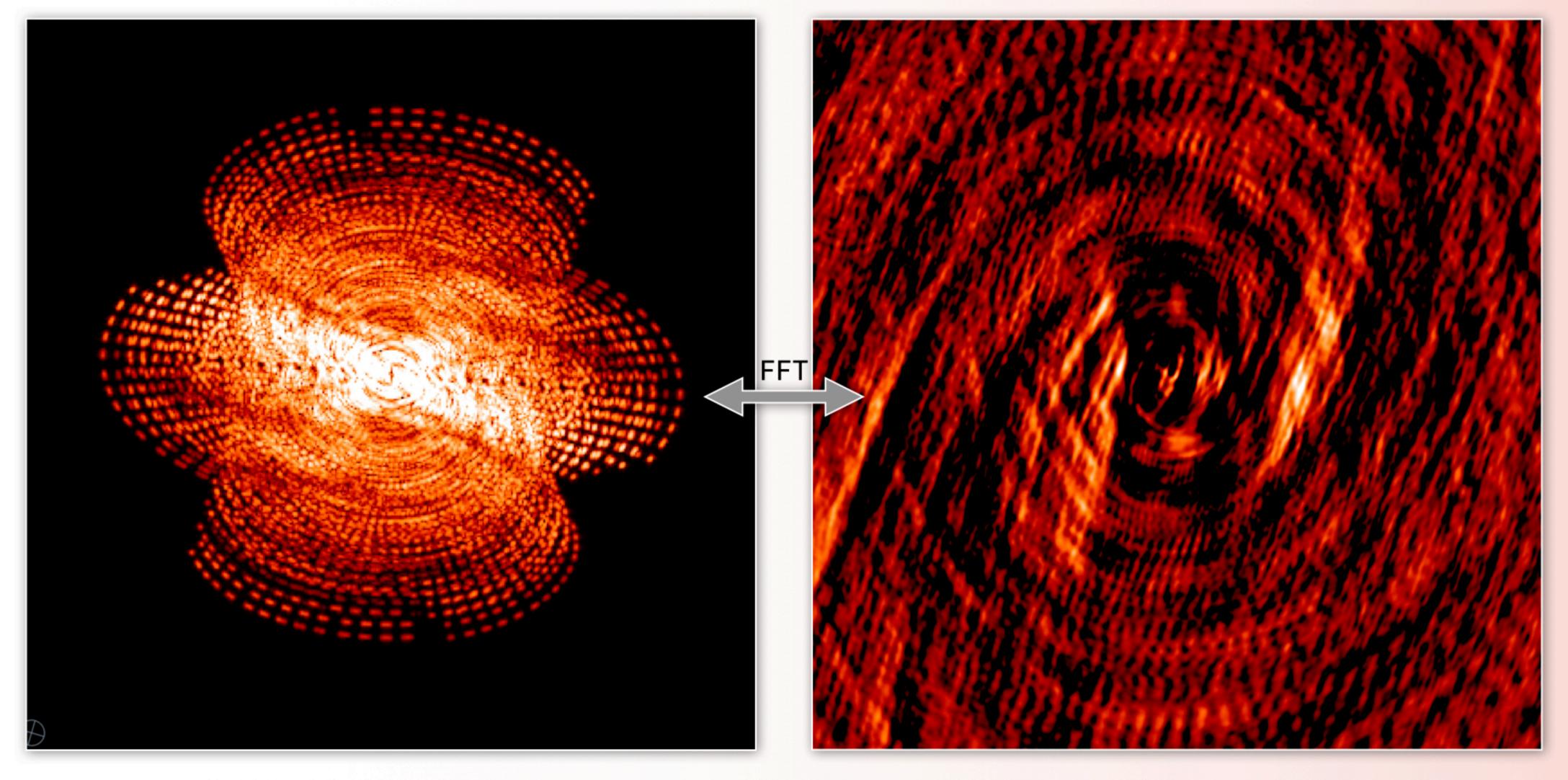
Lots of antennas + earth rotation -> lots of information about the brightness distribution in the Fourier domain

A long track with a modern radio telescope can produce a database containing many billions of visibilities

Gridded visibility amplitudes (uncalibrated)

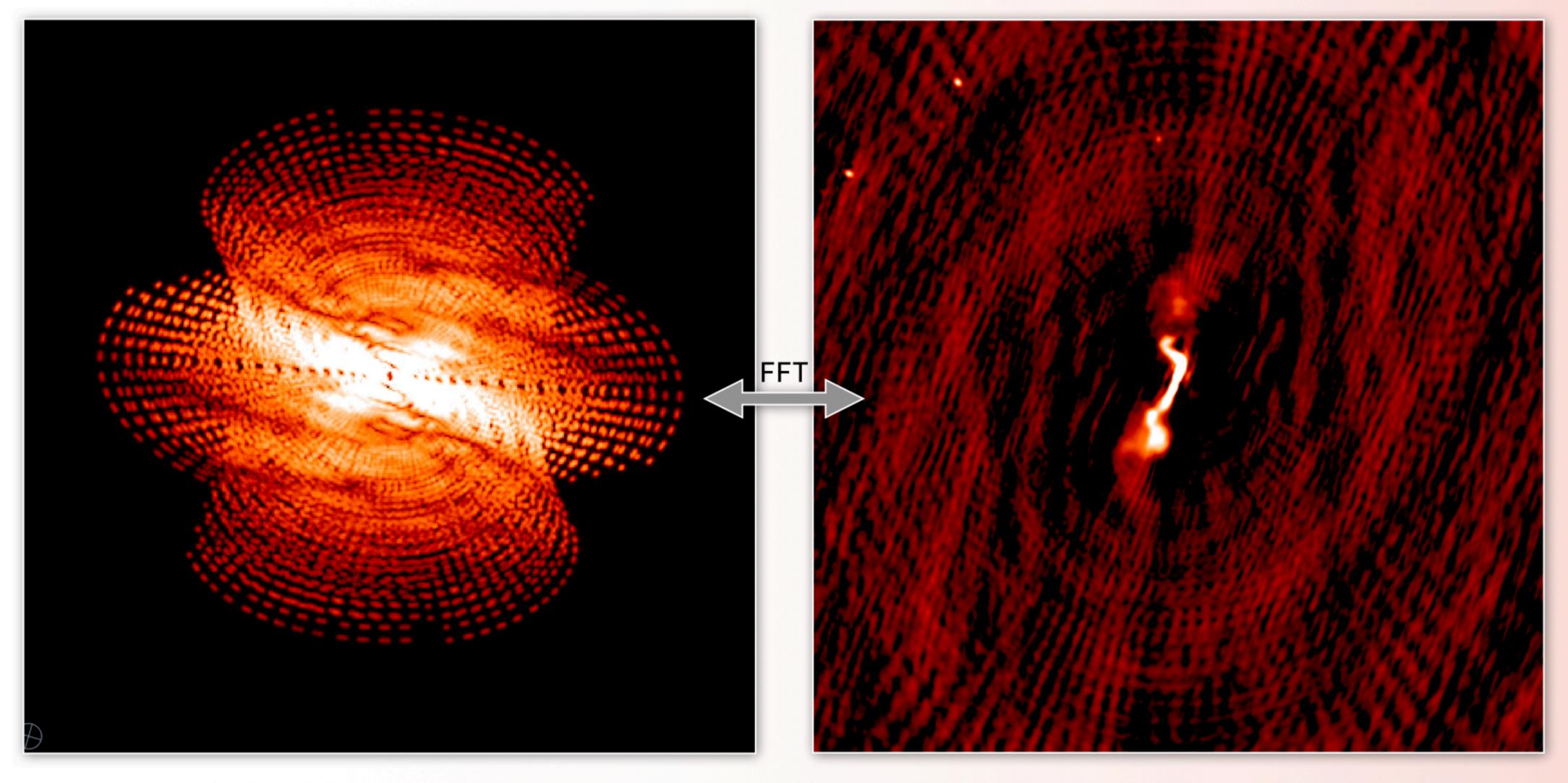


Gridded visibility amplitudes (uncalibrated)



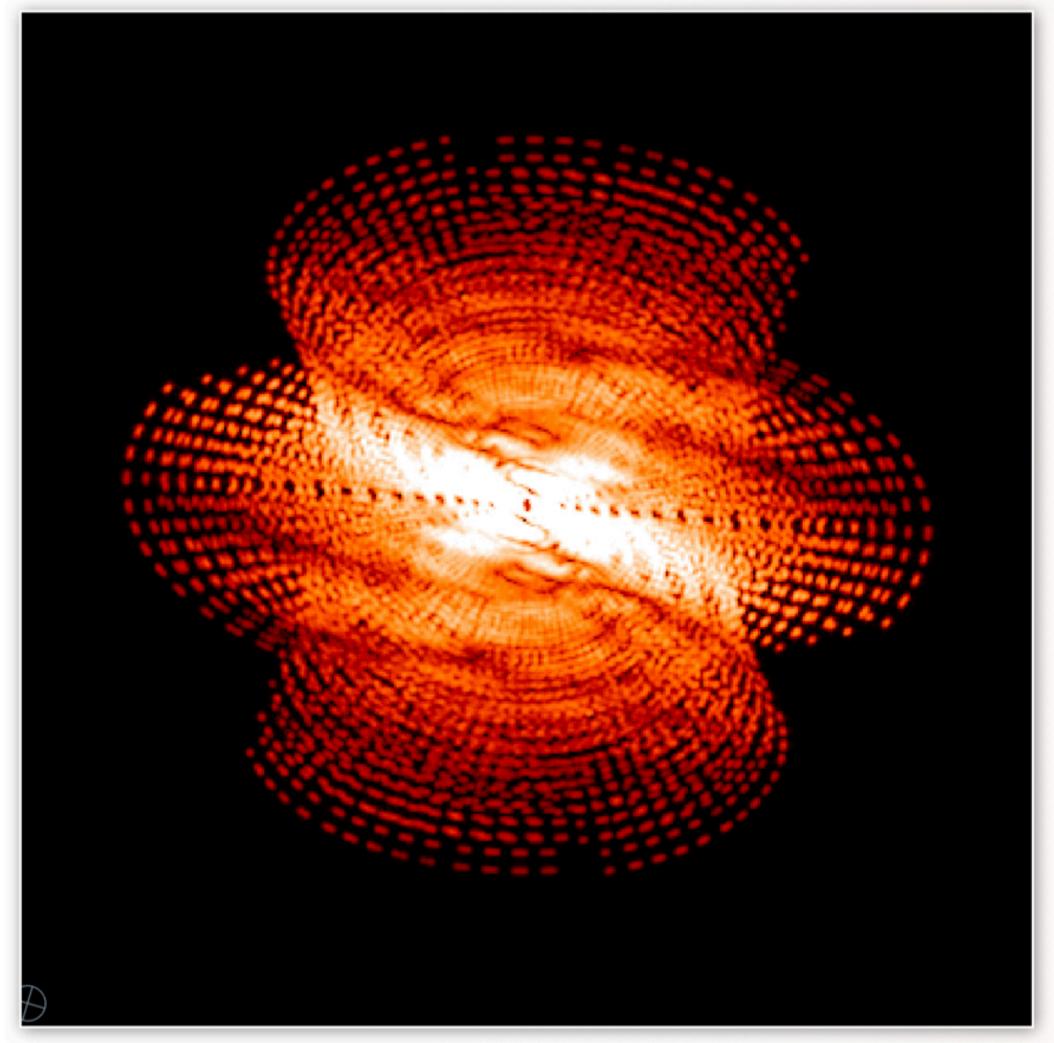
Gridded visibility amplitudes (uncalibrated)

Dirty image (Uncalibrated)

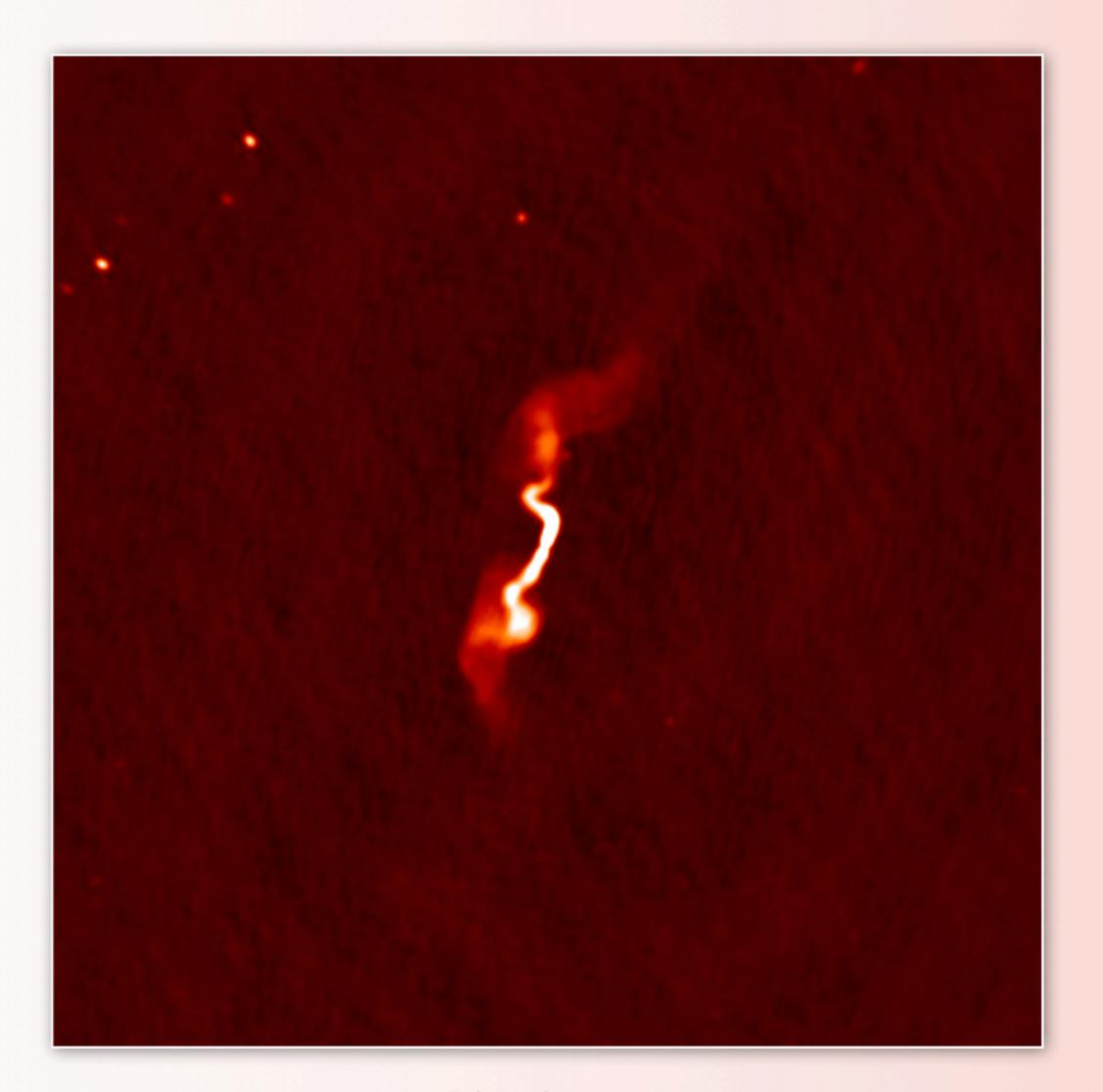


Gridded visibility amplitudes (Calibrated)

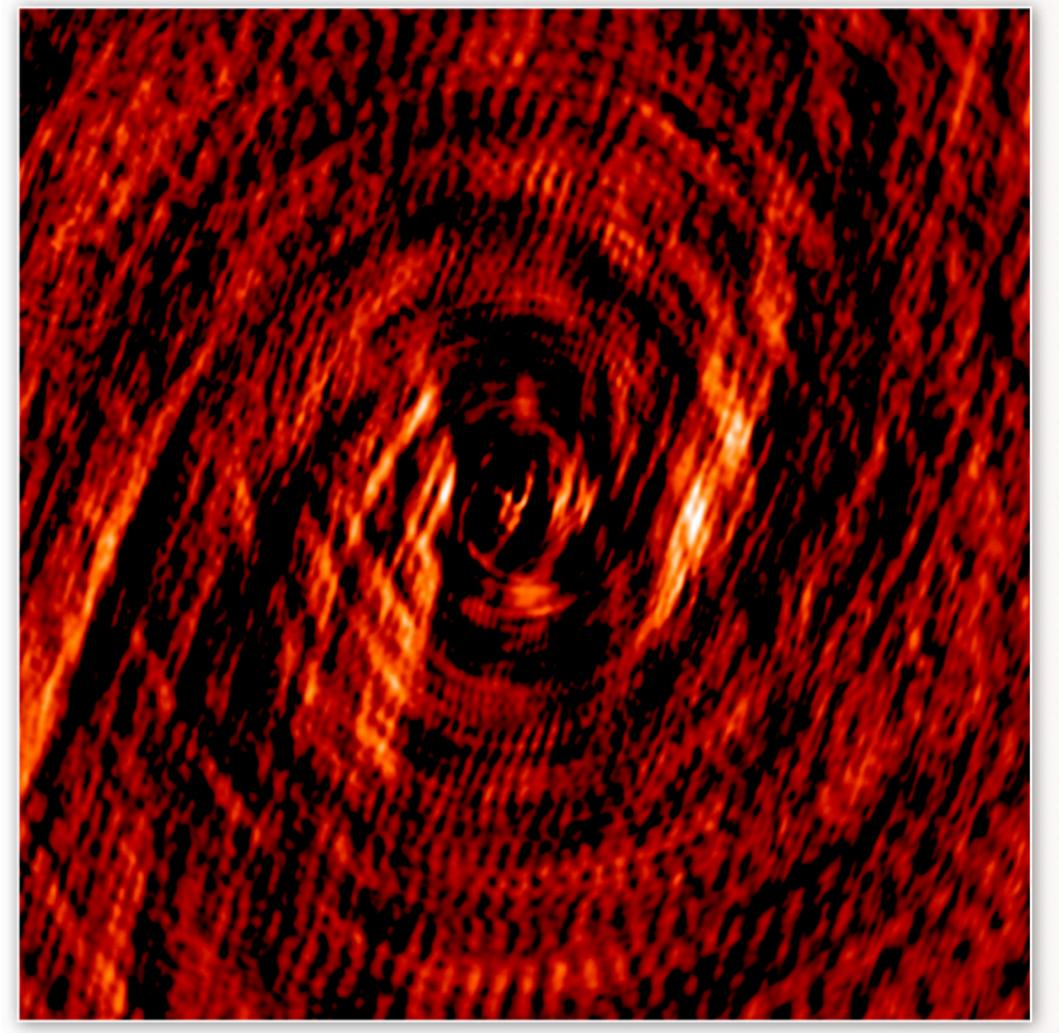
Dirty image (Calibrated)



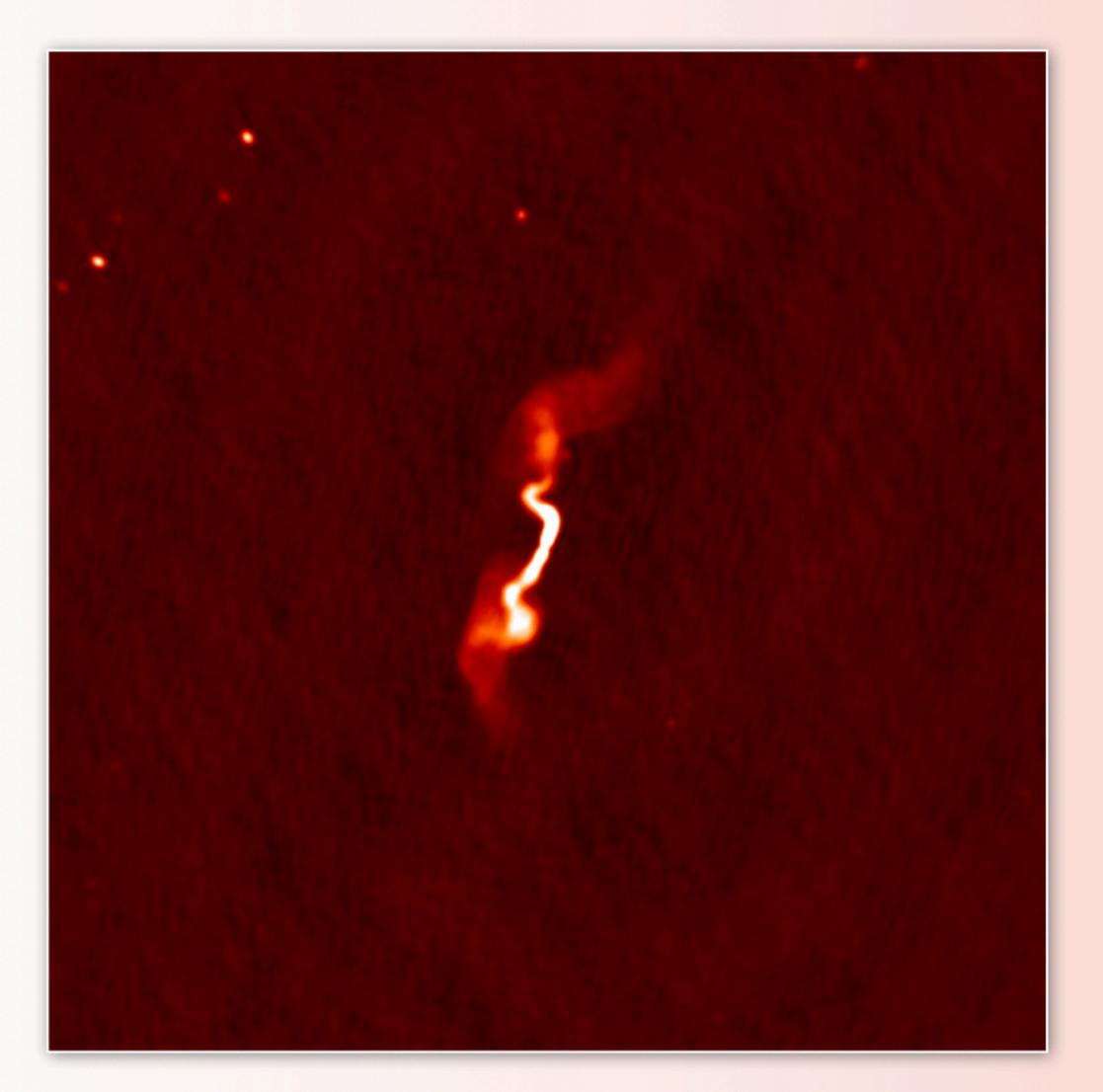
Gridded visibility amplitudes (Calibrated)



Dirty image (Calibrated & Deconvolved)

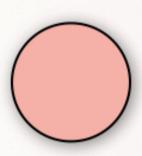


Dirty image (Uncalibrated)

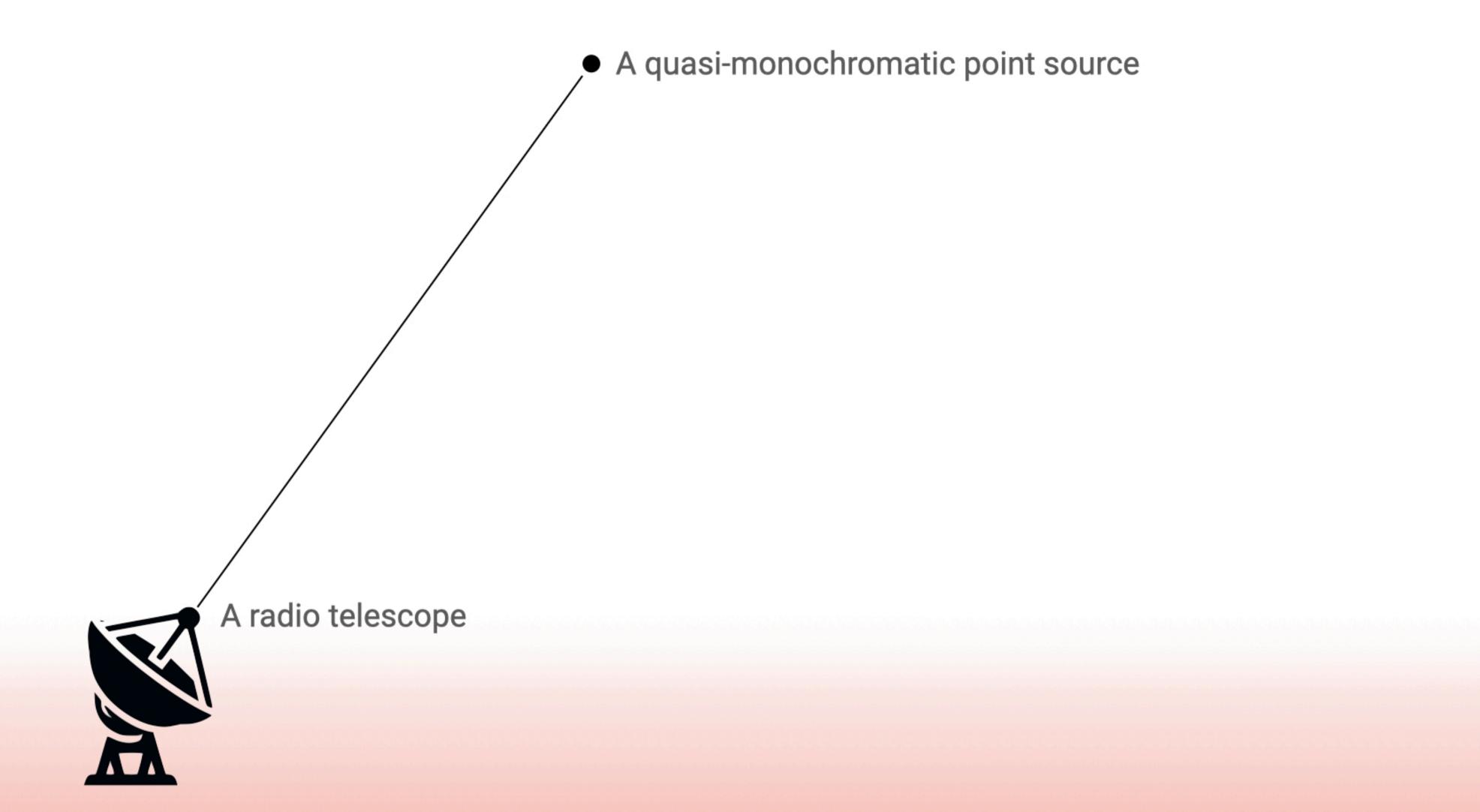


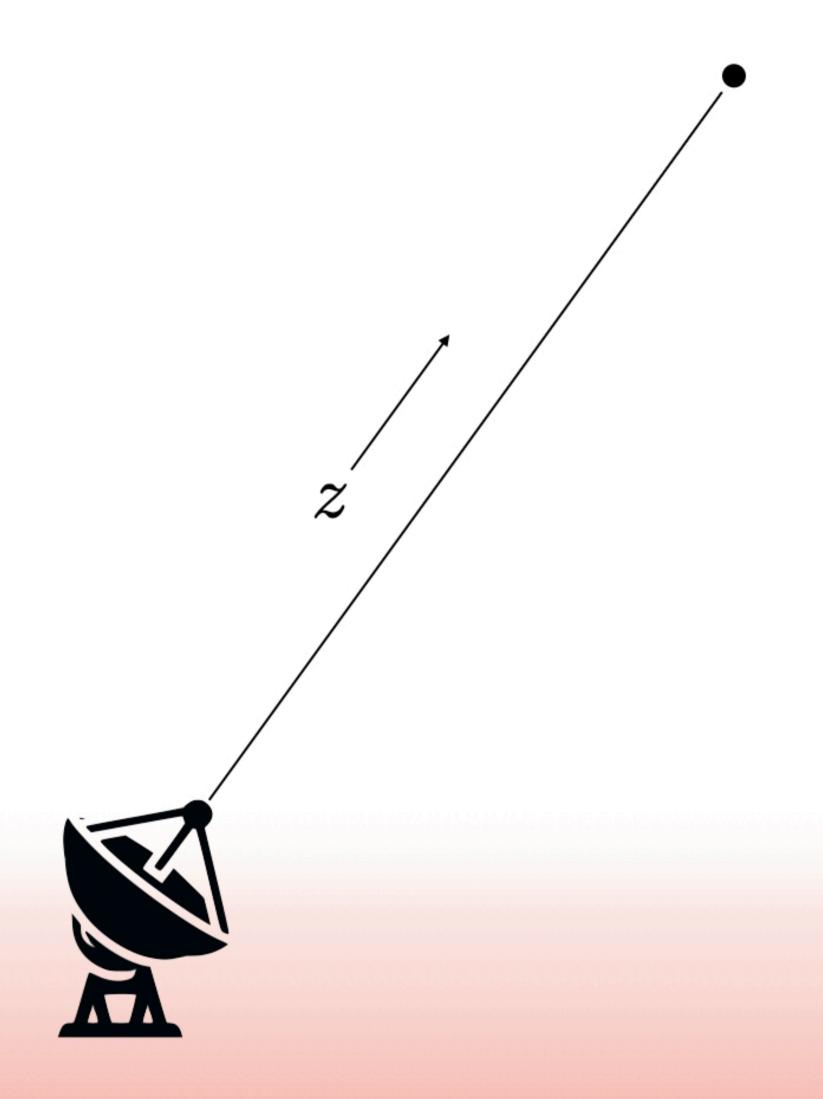
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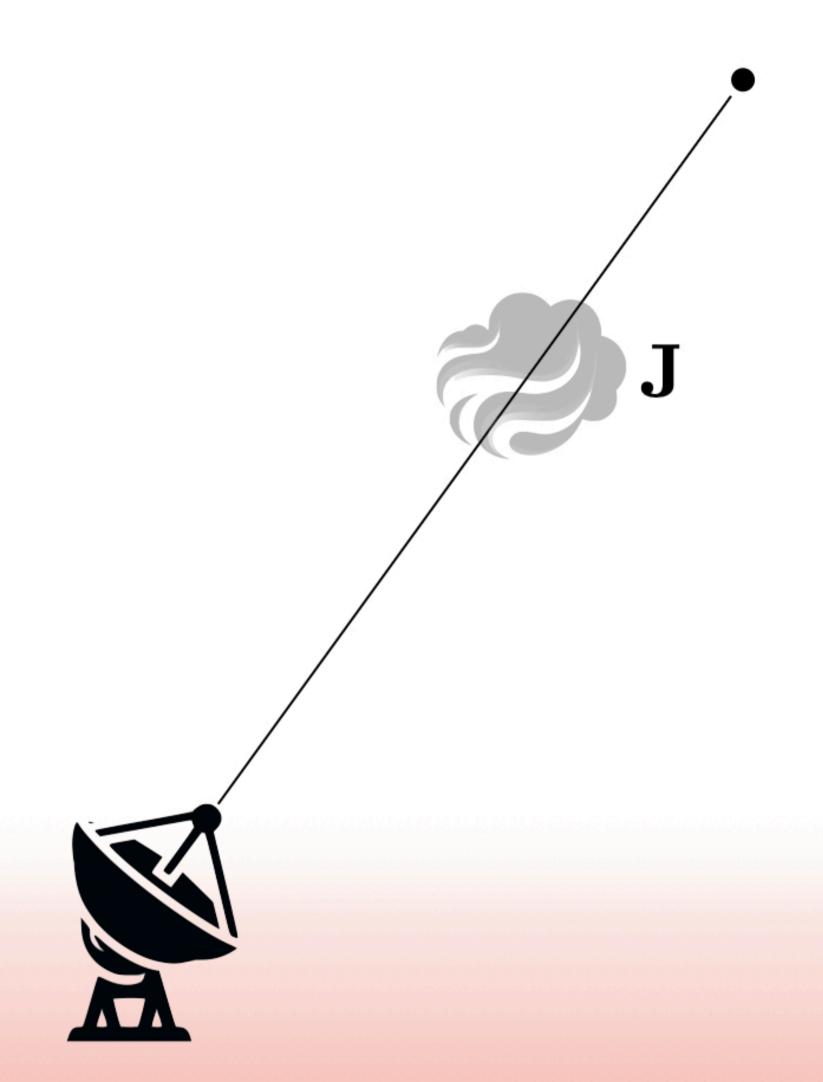
Calibration and the Measurement Equation





The incoming planar electromagnetic wave propagating along direction **z** can be represented by a two-element complex vector, describing the state of the electric field in orthogonal directions **x** and **y**:

$$ar{e} = \begin{vmatrix} e_x \\ e_y \end{vmatrix}$$

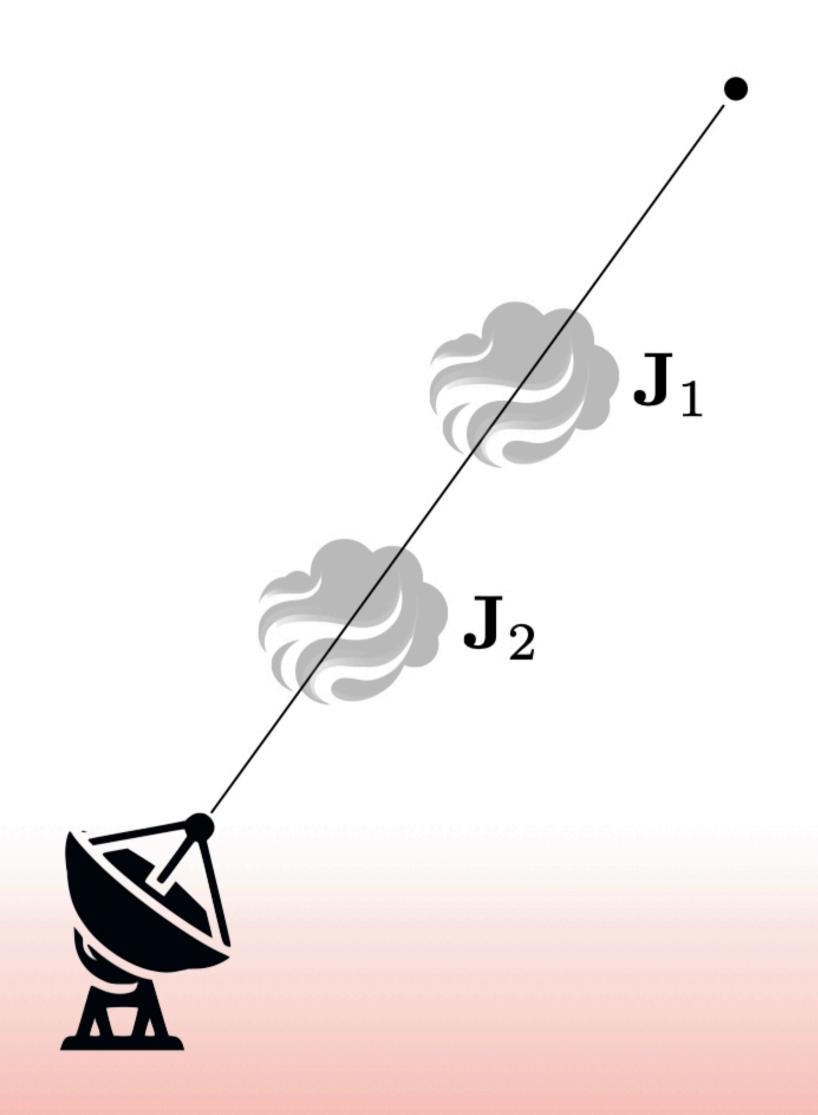


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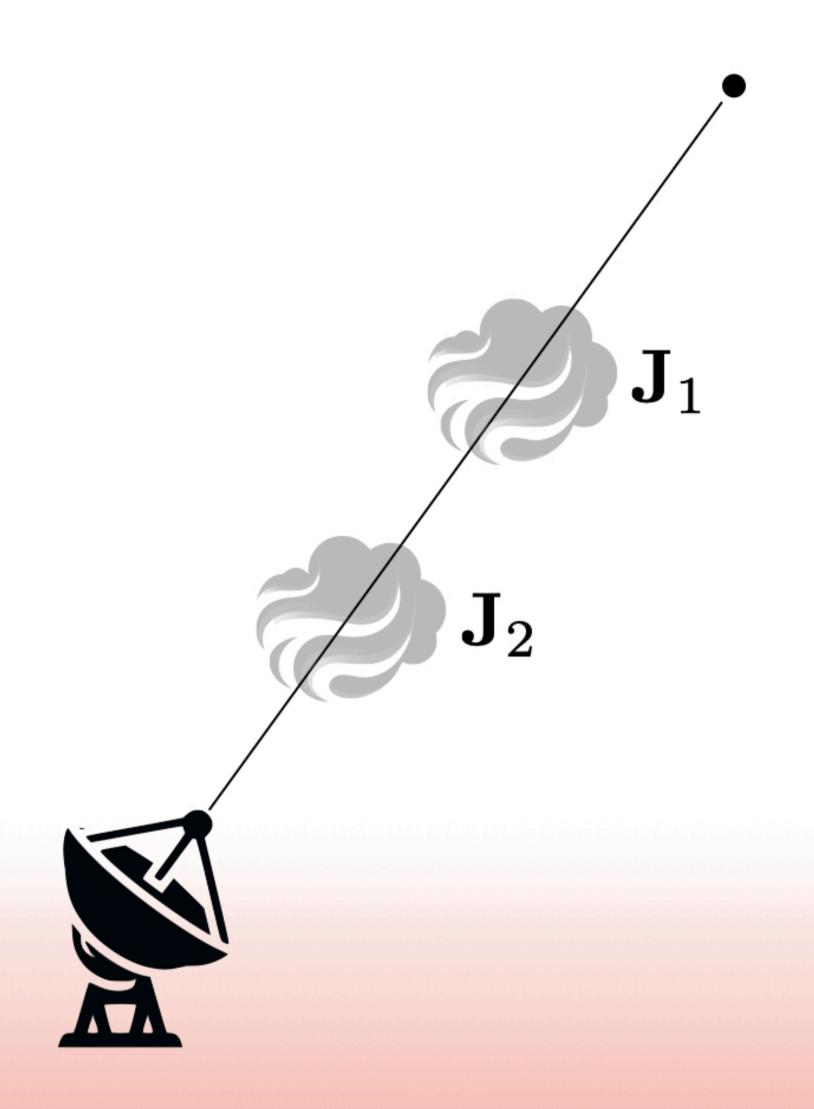
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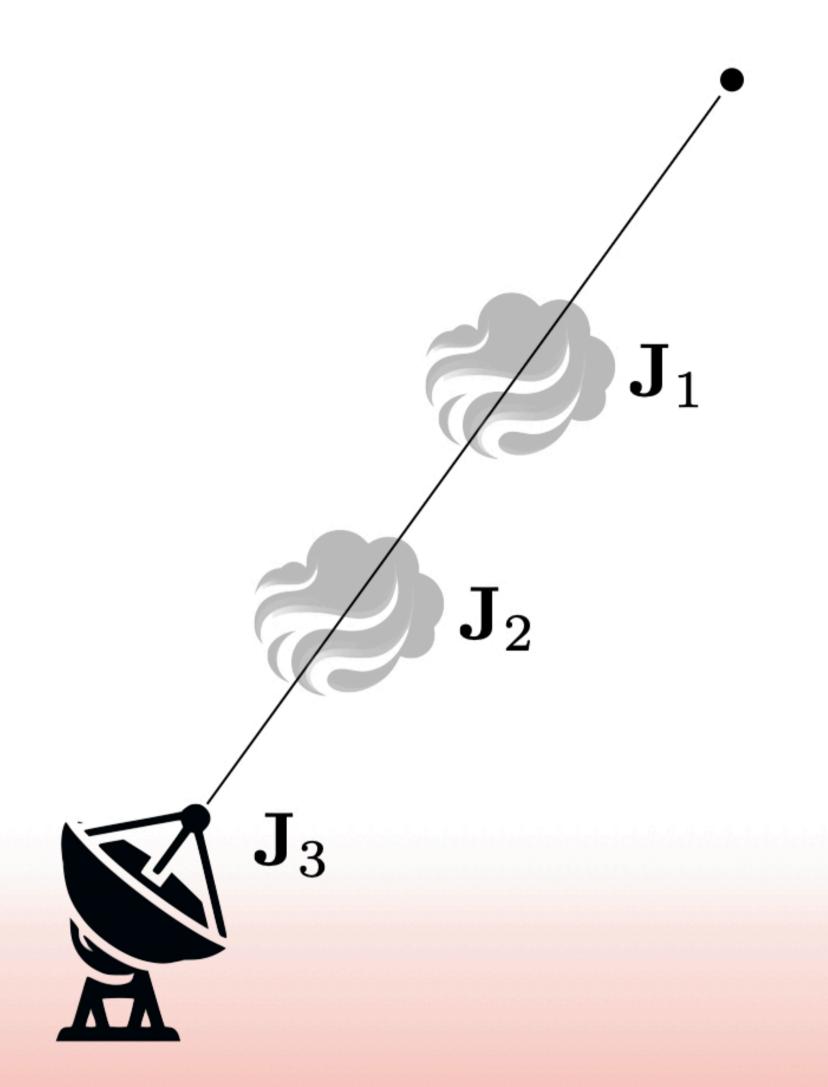
Multiple effects along the path have their own Jones matrix, for which the physical ordering must be preserved:

$$\bar{e}' = \mathbf{J}_n \dots \mathbf{J}_2 \mathbf{J}_1 \bar{e} = \mathbf{J}\bar{e}$$



The incoming radio waves induce complex voltages at the antenna, typically in a pair of feeds  $\boldsymbol{a}$  and  $\boldsymbol{b}$  (which may be orthogonally-aligned linear dipoles, or circular feeds):

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A linear relationship between **v** and **e** can also be represented by a Jones matrix:

$$\bar{v} = \mathbf{J}\bar{e}$$

with **J** being the cumulative product of Jones matrices that relates **v** to **e**, and captures the propagation effects along the signal path.

Two separate antennas *p* and *q* measure independent complex voltage vectors, which relate to the original signal *e* via the two different Jones chains for the two signal paths:

$$ar{v}_p = egin{bmatrix} v_{pa} \ v_{pb} \end{bmatrix} = \mathbf{J}_p egin{bmatrix} e_x \ e_y \end{bmatrix}$$

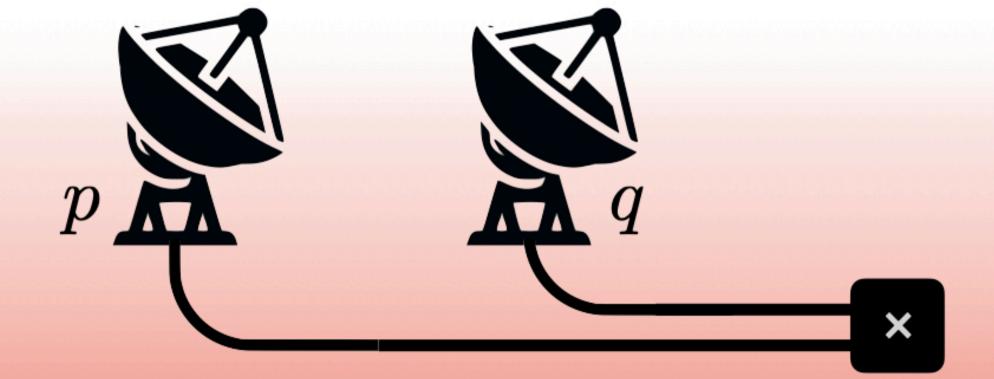
$$ar{v}_q = egin{bmatrix} v_{qa} \ v_{qb} \end{bmatrix} = \mathbf{J}_q egin{bmatrix} e_x \ e_y \end{bmatrix}$$





Radio interferometers work by correlating the voltages from antennas p and q, averaged over some small interval, equivalent to forming the outer product of voltage vectors  $\mathbf{v}_p$  and  $\mathbf{v}_q$ :

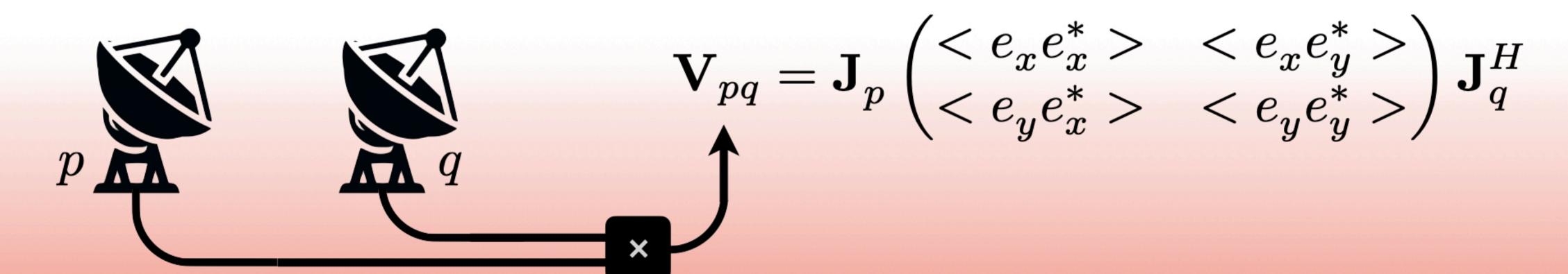
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The **complex visibility**  $V_{pq}$  is thus a measurement related to the incoming electromagnetic wave  $\mathbf{e}$ , corrupted by the total Jones chains along the differing paths to antennas p and q, assuming the Jones terms are constant across the averaging interval:



$$\mathbf{V}_{pq} = \mathbf{J}_{p} \begin{pmatrix} < e_{x}e_{x}^{*} > & < e_{x}e_{y}^{*} > \\ < e_{y}e_{x}^{*} > & < e_{y}e_{y}^{*} > \end{pmatrix} \mathbf{J}_{q}^{H}$$





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can be expressed in terms of four Stokes parameters:

$$\mathbf{V}_{pq} = \mathbf{J}_p \begin{pmatrix} I + Q & U + iV \\ U - iV & I - Q \end{pmatrix} \mathbf{J}_q^H$$





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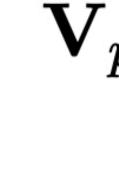
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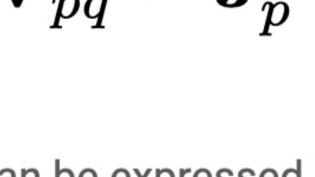
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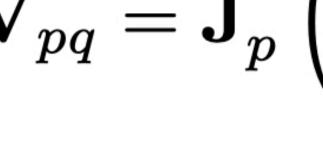
The complex visibility measurements thus give us information about the polarised radio sky, provided we can understand the Jones matrices.



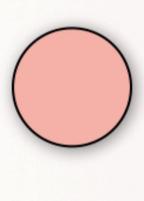








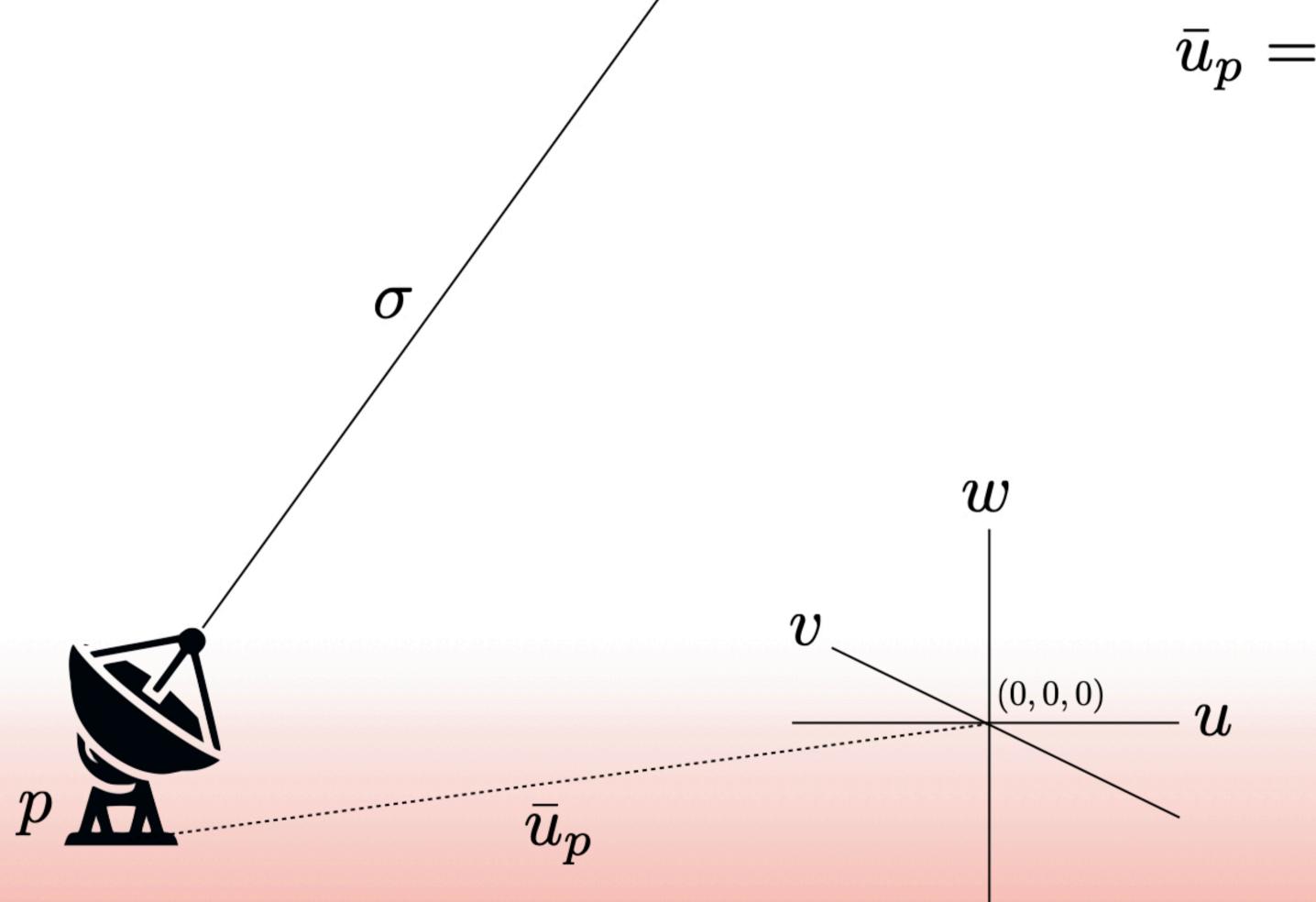




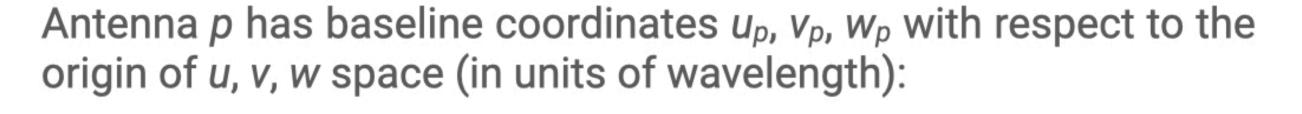
Jones Matrices

Antenna p has baseline coordinates  $u_p$ ,  $v_p$ ,  $w_p$  with respect to the origin of u, v, w space (in units of wavelength):

$$\bar{u}_p = (u_p, v_p, w_p)$$



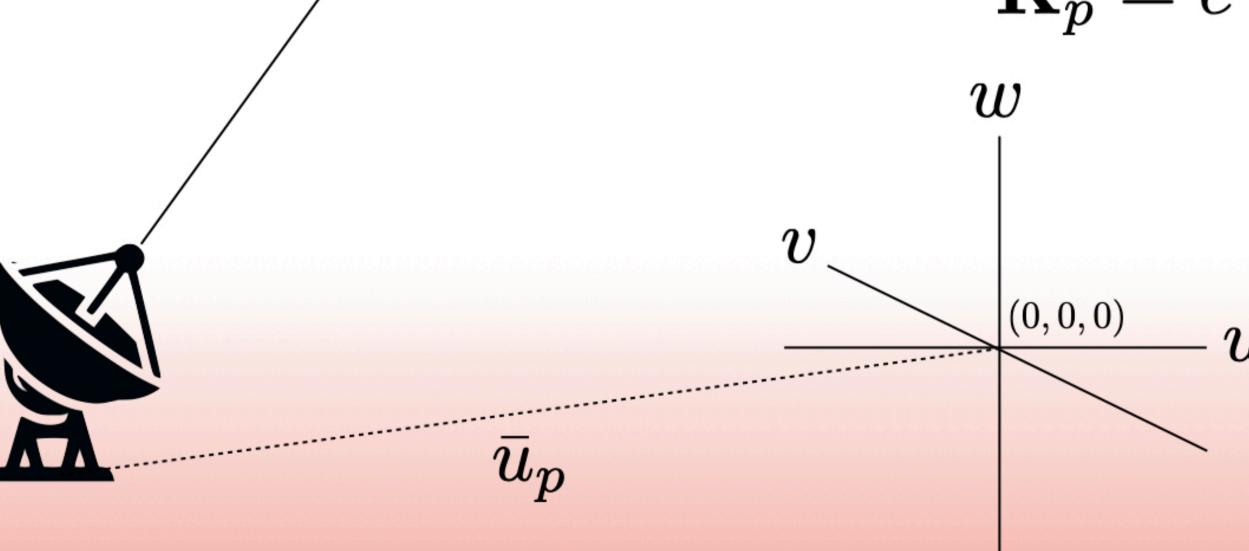
 $\sigma$ 

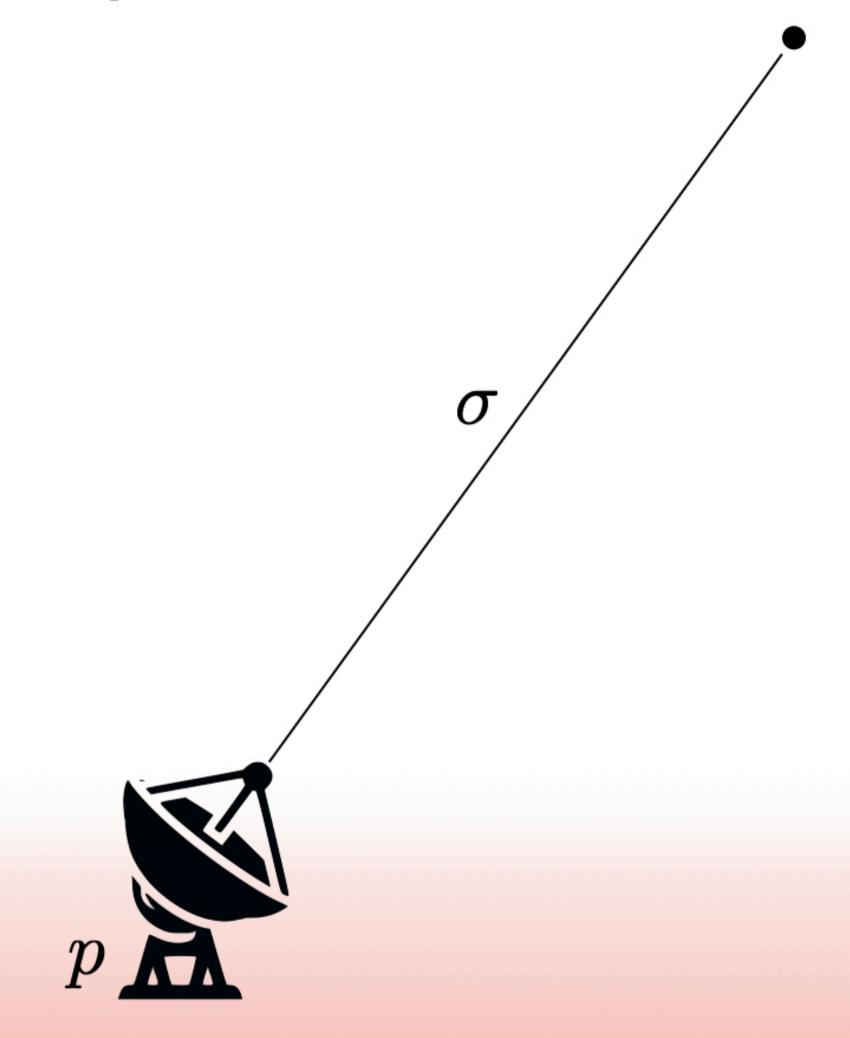


$$\bar{u}_p = (u_p, v_p, w_p)$$

A signal arriving from direction  $\sigma$  with direction cosines l, m, n will have a phase offset that can be expressed as a scalar Jones matrix:

$$\mathbf{K}_p = e^{-2\pi i(u_p l + v_p m + w_p(n-1))}$$





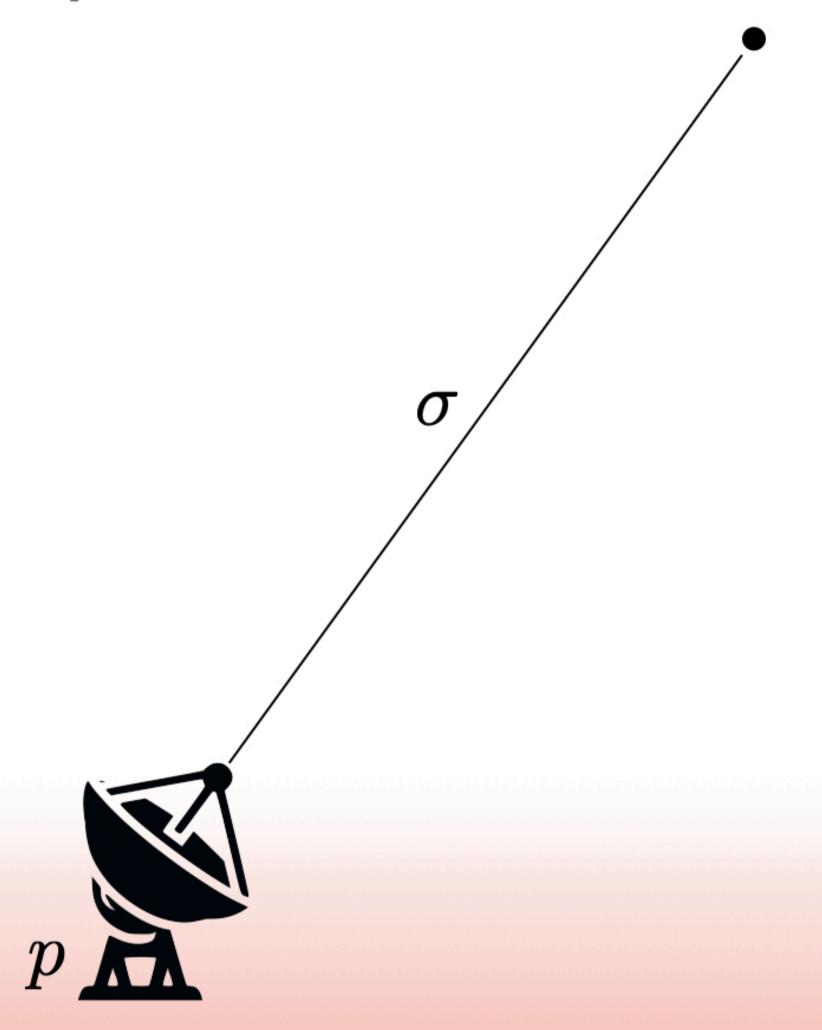
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In practice, the correlator introduces additional delay contributions in order to track a particular direction of interest on the sky.

This is known as the phase centre and is usually (but doesn't have to be) coincident with the direction that the antennas are tracking.

Our Measurement Equation for the baseline p, q now becomes:

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or by defining the "source coherency"  $X_{pq}$  as the product of the brightness matrix and the geometric delay terms for p and q, we obtain:

$$\mathbf{V}_{pq} = \mathbf{J}_p X_{pq} \mathbf{J}_q^H$$

for a baseline tracking a single point source corrupted by the Jones chains  $J_p$  and  $J_q$ .





#### Links in the Jones chain

$$\mathbf{J}_p = \mathbf{K}_p \mathbf{B}_p \mathbf{G}_p \mathbf{D}_p \mathbf{E}_p \mathbf{X}_p \mathbf{P}_p \mathbf{T}_p \mathbf{Z}_p$$

 $\mathbf{K}_{p}$  geometric delay

 $\mathbf{B}_p$  bandpass

 $\mathbf{G}_{p}$  electronic gains

 $\mathbf{D}_{p}$  polarization leakage

 $\mathbf{E}_{p}$  antenna primary beam response

 $\mathbf{X}_p$  cross-hand phases

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Fundamental domains are:

$$\mathbf{E}_p = \mathbf{E}_p(l, m, \nu, t)$$

but we often assume that certain Jones matrices are only strong functions or a reduced set of these, e.g.:

$$\mathbf{B}_p = \mathbf{B}_p(\nu)$$

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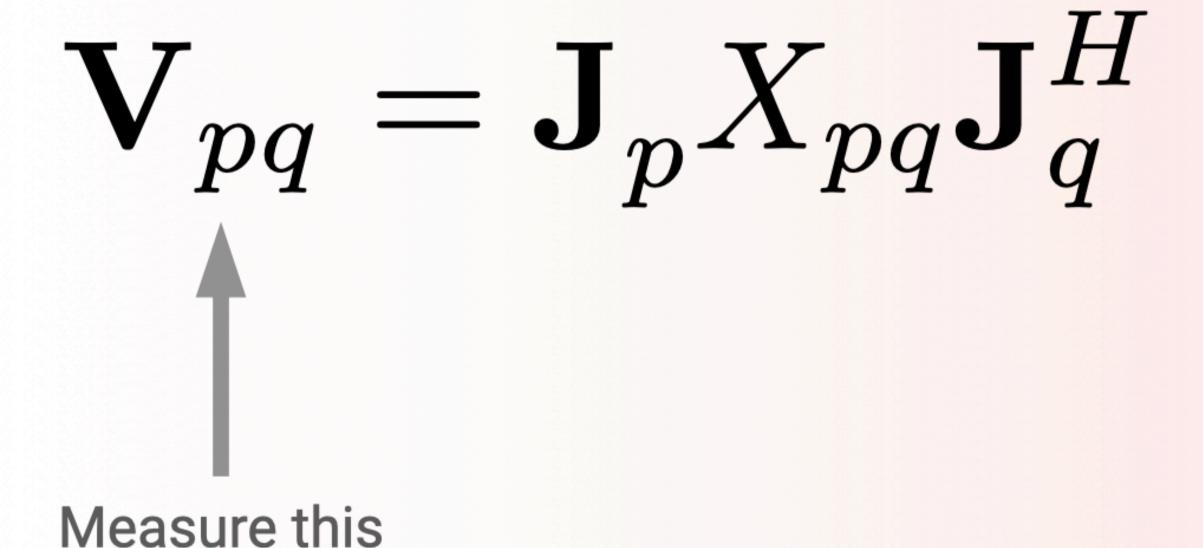
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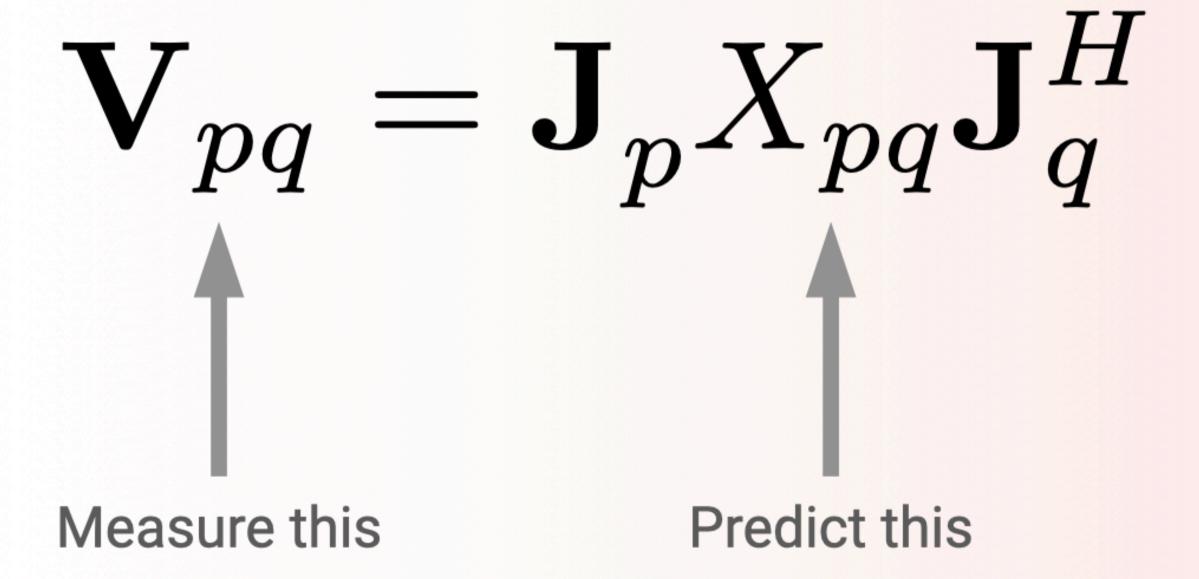
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Some of these will feature in more detail in the **Advanced Calibration** and **Polarisation** lectures.

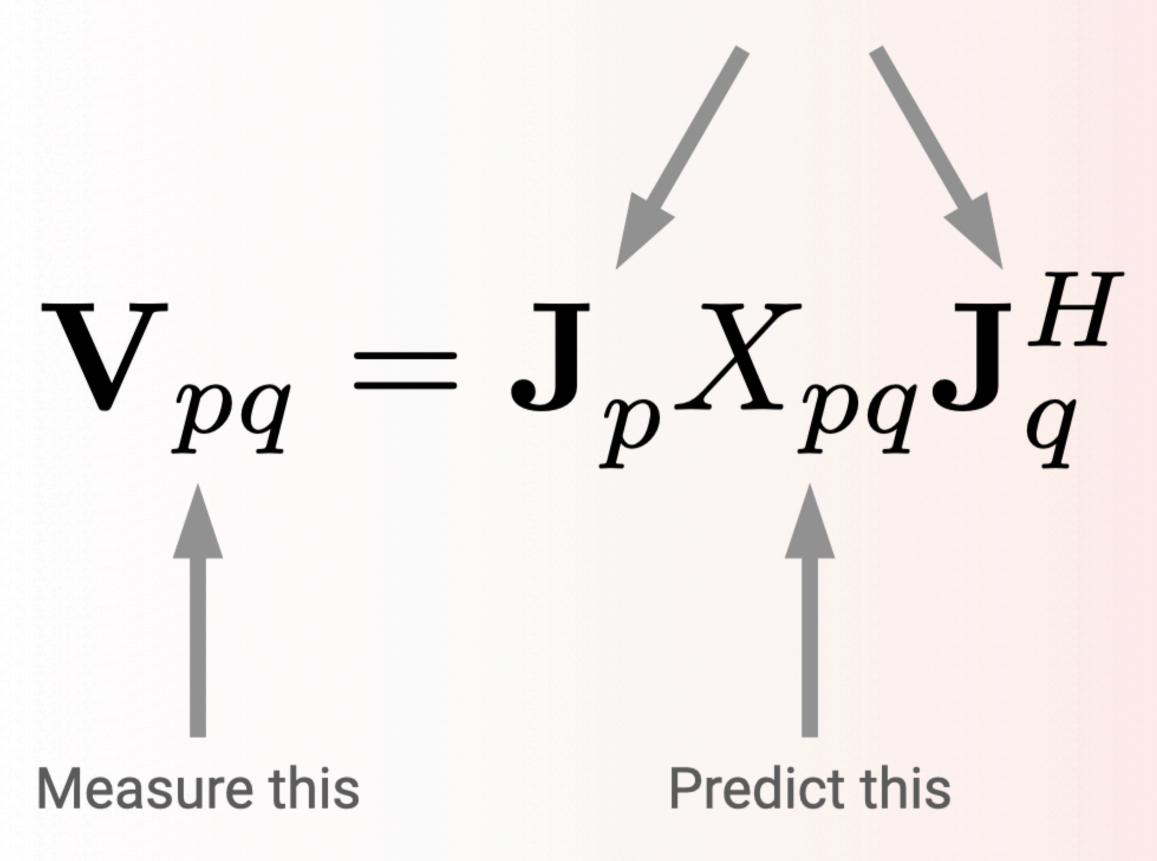
I subjected you to all that because (IMO) calibration is best understood by this mathematical formalism, known as the Measurement Equation

$$\mathbf{V}_{pq} = \mathbf{J}_p X_{pq} \mathbf{J}_q^H$$





Solve for these



$$\min_{\mathbf{G}} ||\mathbf{D} - \mathbf{GMG}^H||$$

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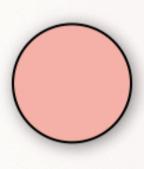
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There are two tiers of calibrators: **primary** and **secondary**. Their roles are different depending on what instrumental correction we are trying to derive. The strategy for observing them will depend on the characteristics of the array, the observing frequency, and what the demands of your science goals are.



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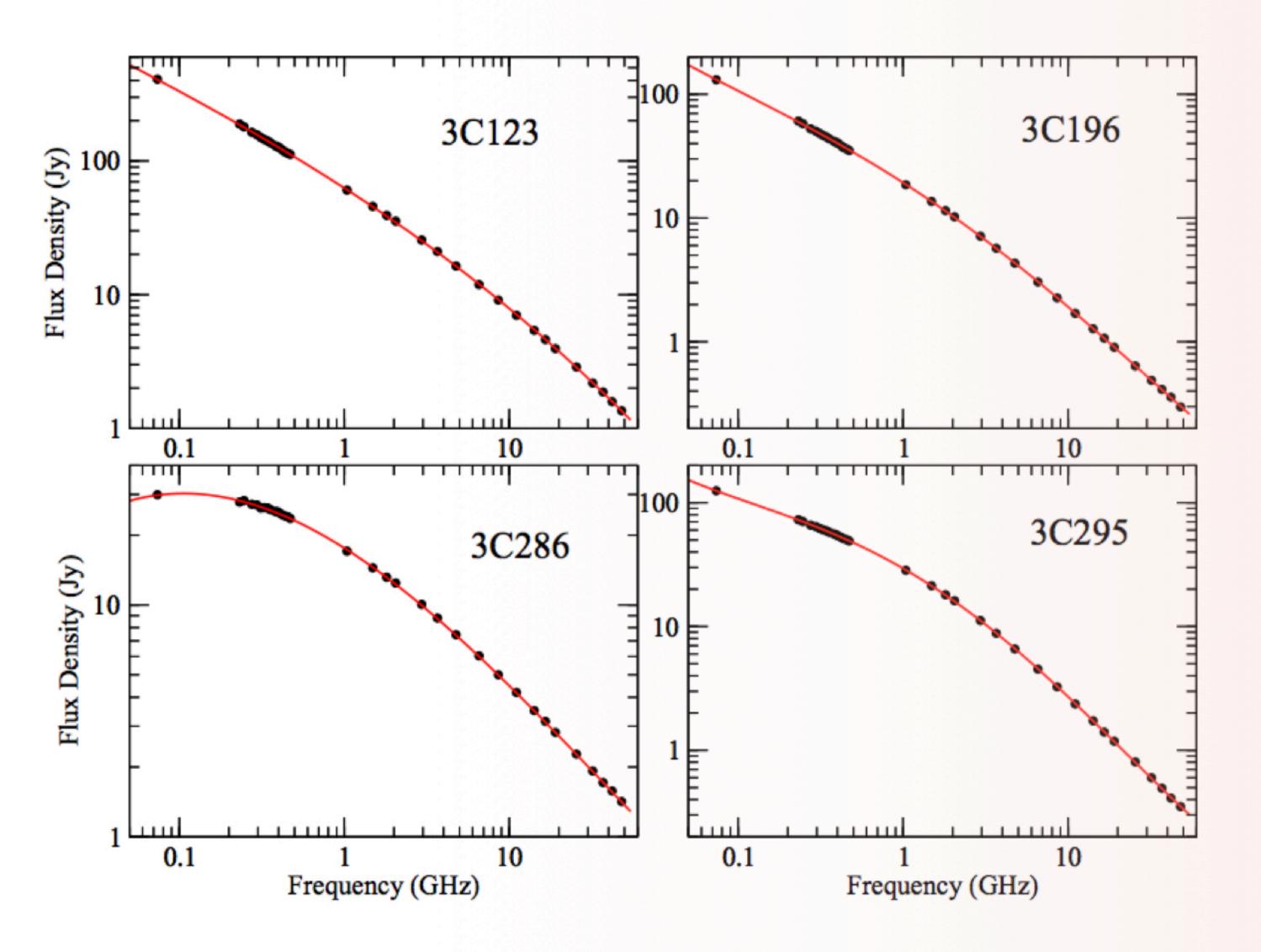
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- Correct the instrumental response to polarised radio waves: the polarisation properties of the calibrator source must be precisely known (more on this in the polarimetry lecture).



Measured and modelled spectra of four 3C calibrator sources.

These four are stable to better than 1% over a 20 year timescale.



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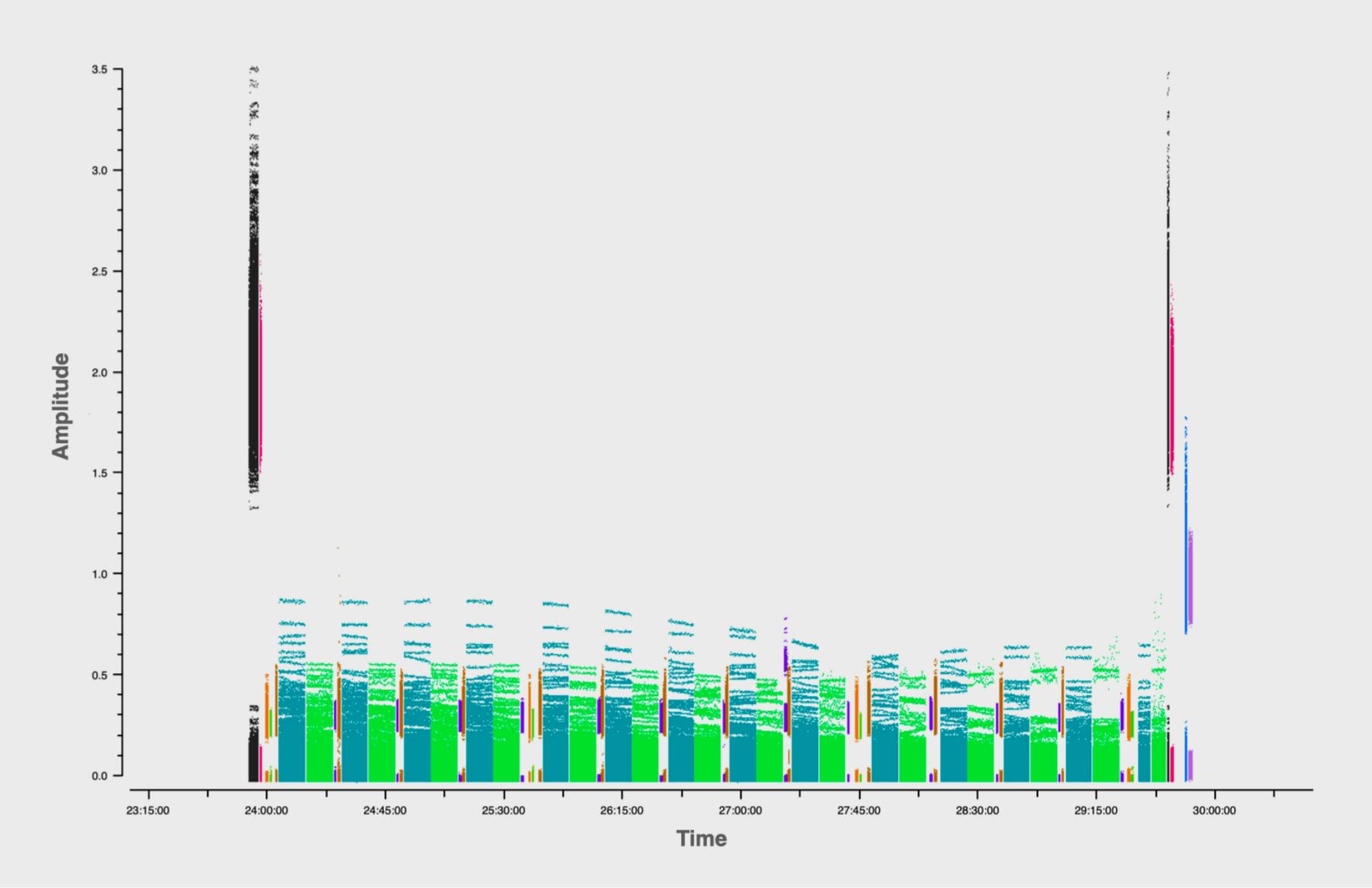
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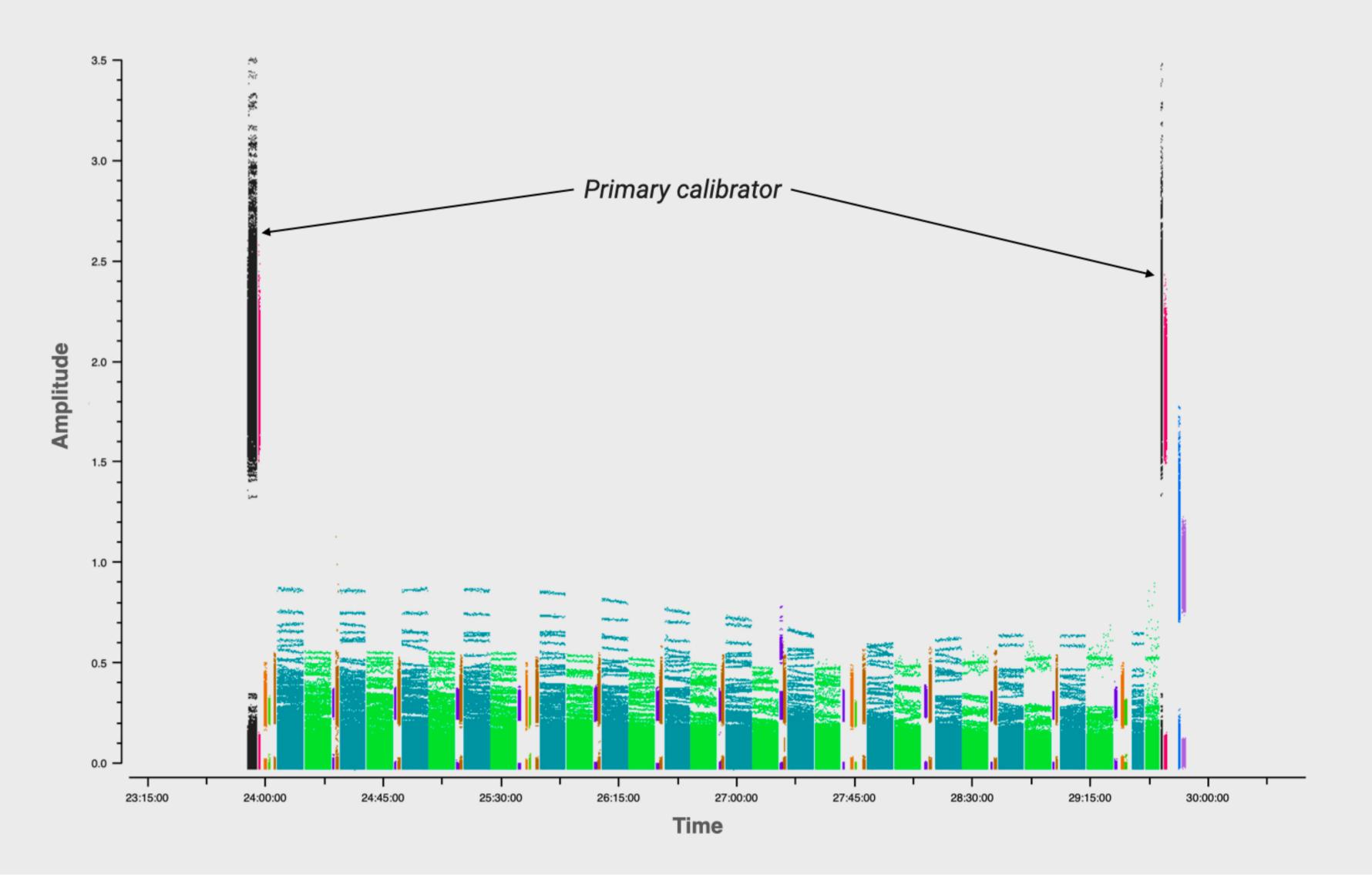
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- With this method we leverage the brightness and compactness of the secondary, and scale the amplitudes
  of its gain corrections to those of the well-modelled primary to avoid any brightness ambiguity.

- Calibrators in the second tier are much more numerous, but generally less well monitored.
- Secondary calibrators are generally selected to be close to the target, and are used to derive timedependent instrumental corrections (e.g. complex antenna gains, parallactic angle corrections for polarimetry).
- Ideally these also have VLBI-grade positional measurements to preserve the astrometric accuracy.
- How often the secondary is visited depends on array layout, frequency, and observing conditions.
- The secondary may or may not exhibit variability. This is guarded against by using a unity-amplitude point source visibility model.
- With this method we leverage the brightness and compactness of the secondary, and scale the amplitudes
  of its gain corrections to those of the well-modelled primary to avoid any brightness ambiguity.
- Be wary of the spectral axis at this stage. SPWs in VLA / ALMA data will naturally handle it, this might not be the case for other instruments!

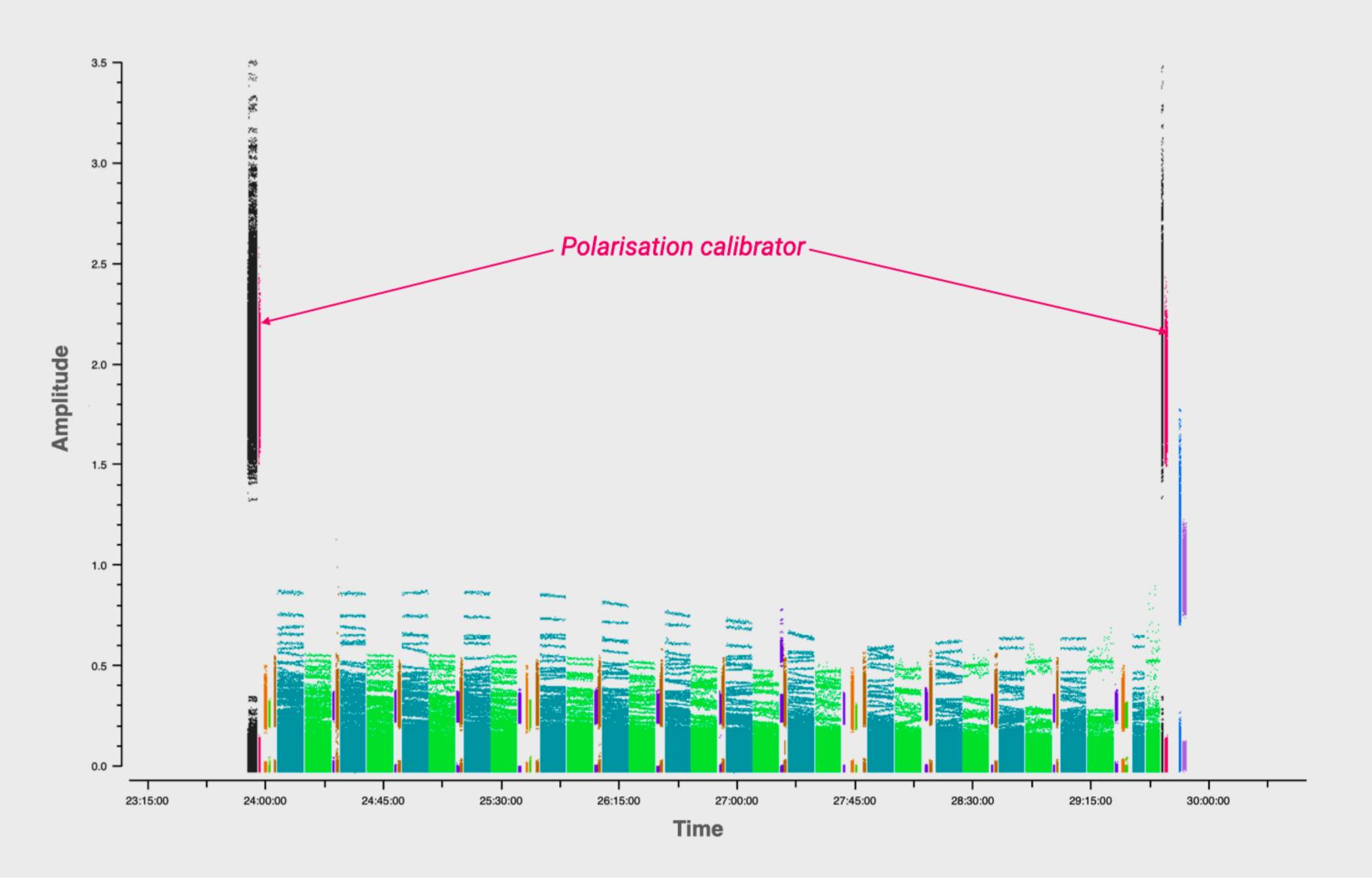
# Anatomy of a scheduling block



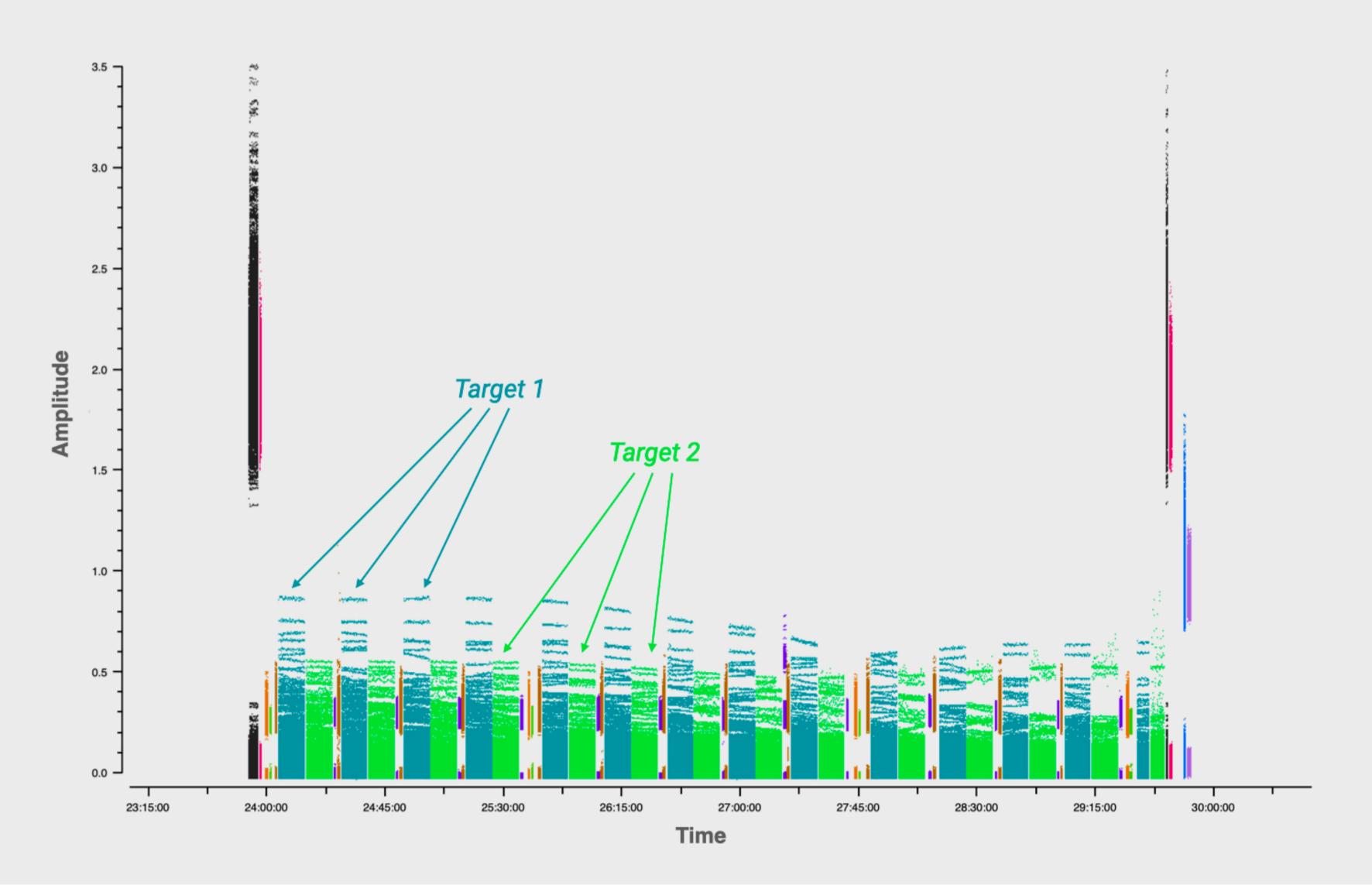
# Anatomy of a scheduling block



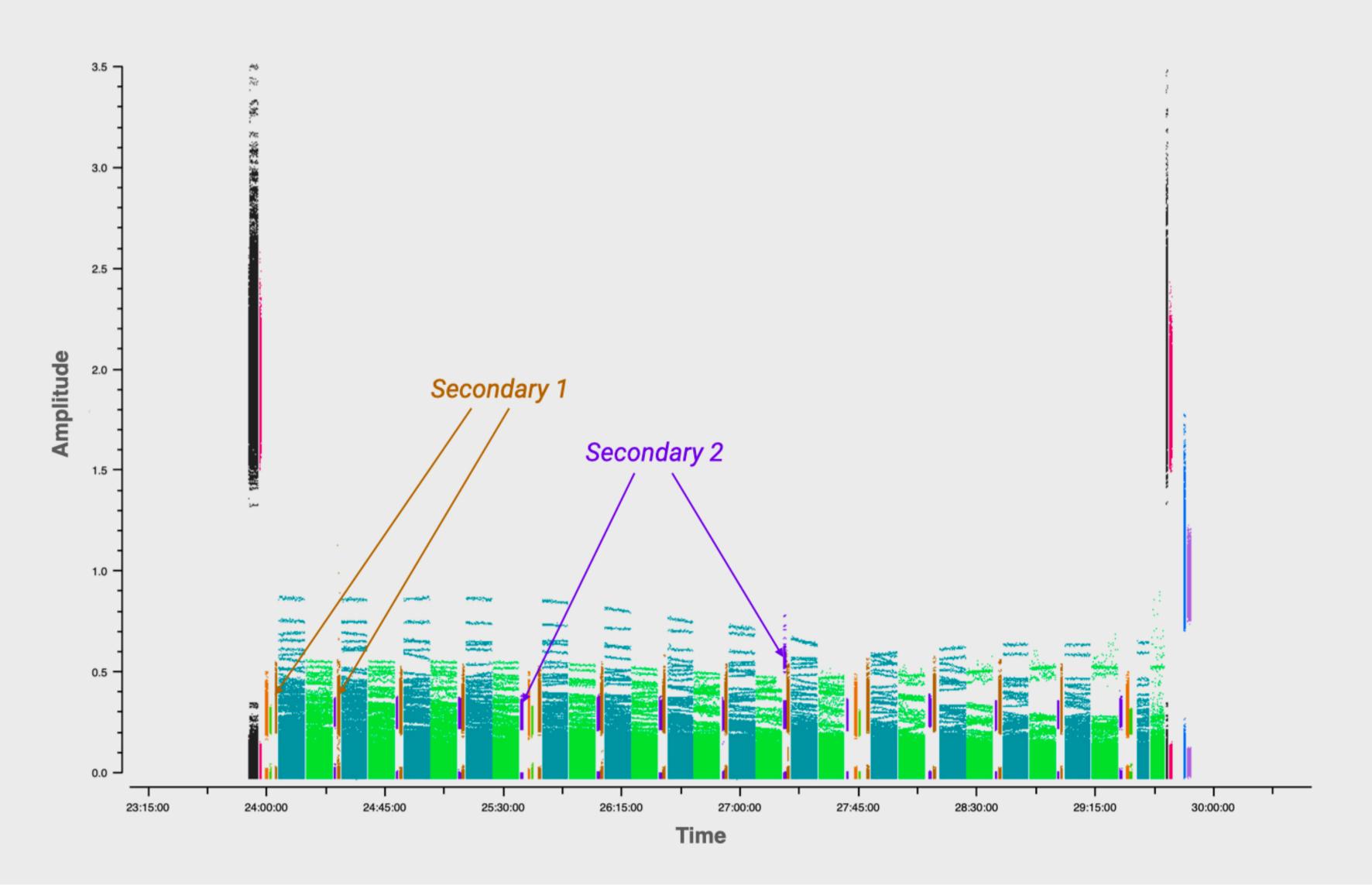
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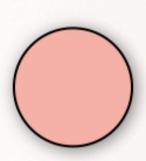


# Anatomy of a scheduling block



# Anatomy of a scheduling block





A Calibration Example

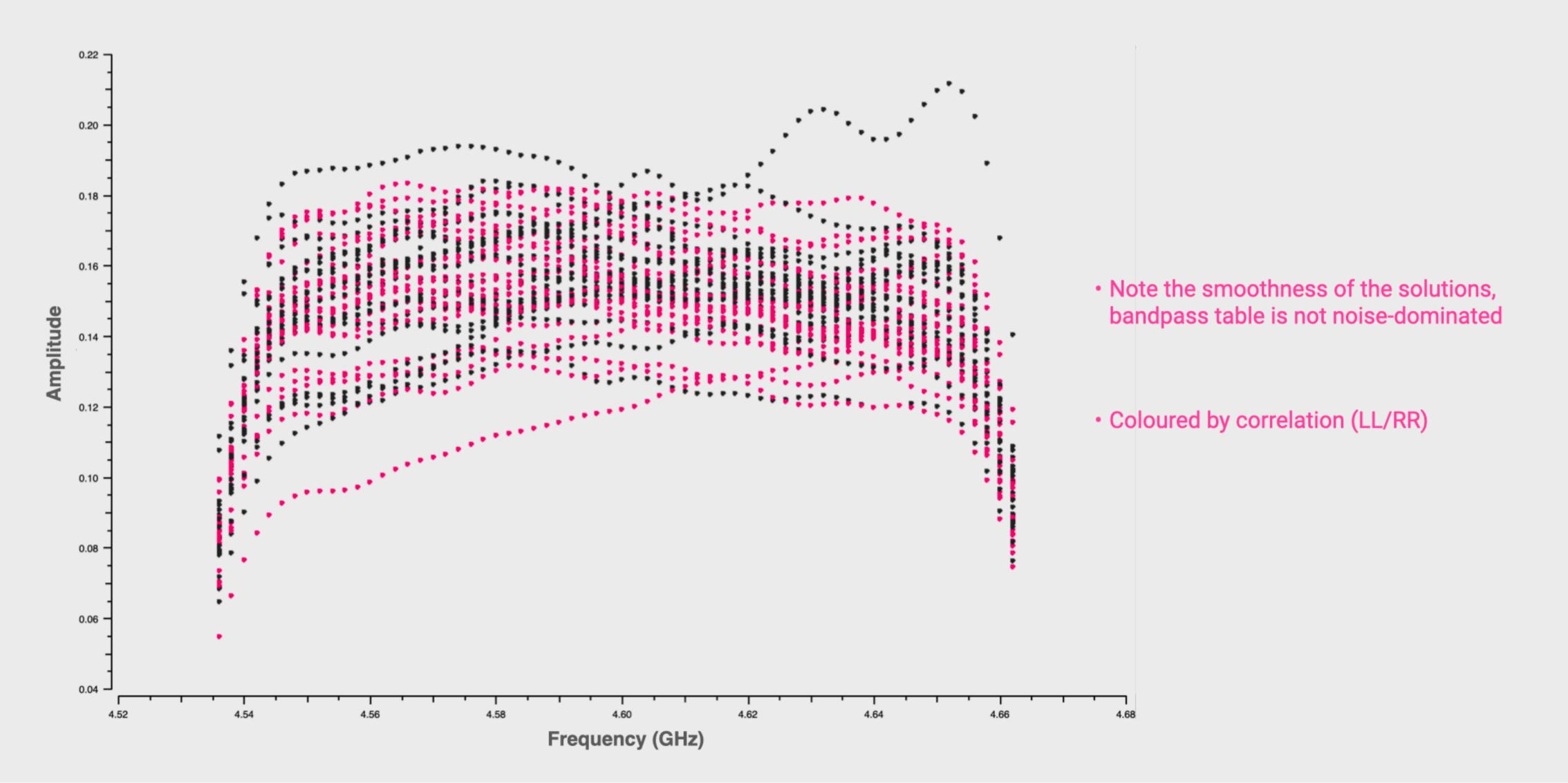


- Demonstrating only a subset of the full observation:
  - 'C1' pointing only: RA = 18h49m24.24s, Dec = -00d55m40.58s
  - Single spectral window: 4.6 GHz, 64 frequency channels, 128 MHz bandwidth
  - About 1 hour on-target

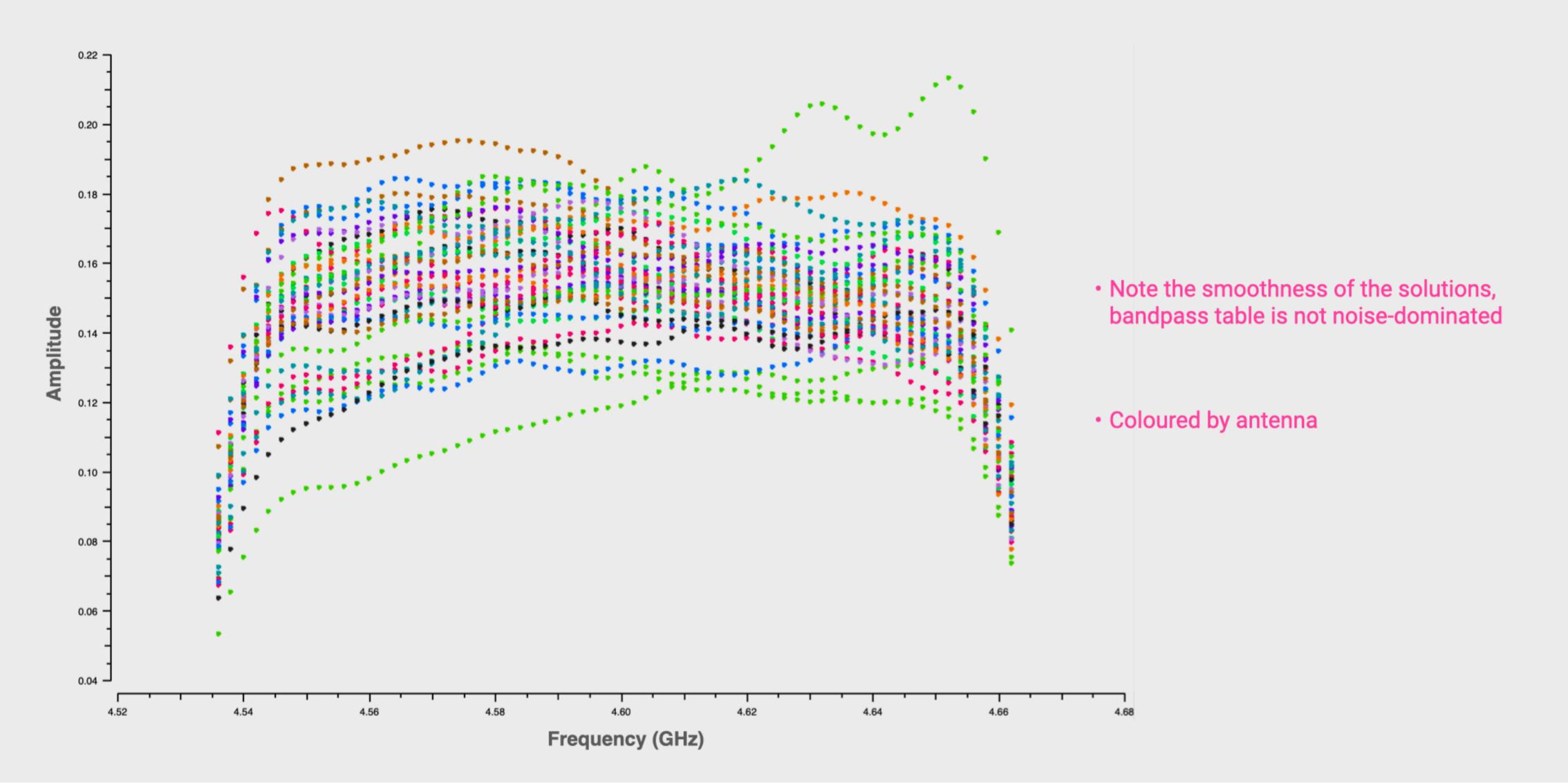
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- Try it yourself: https://casaguides.nrao.edu

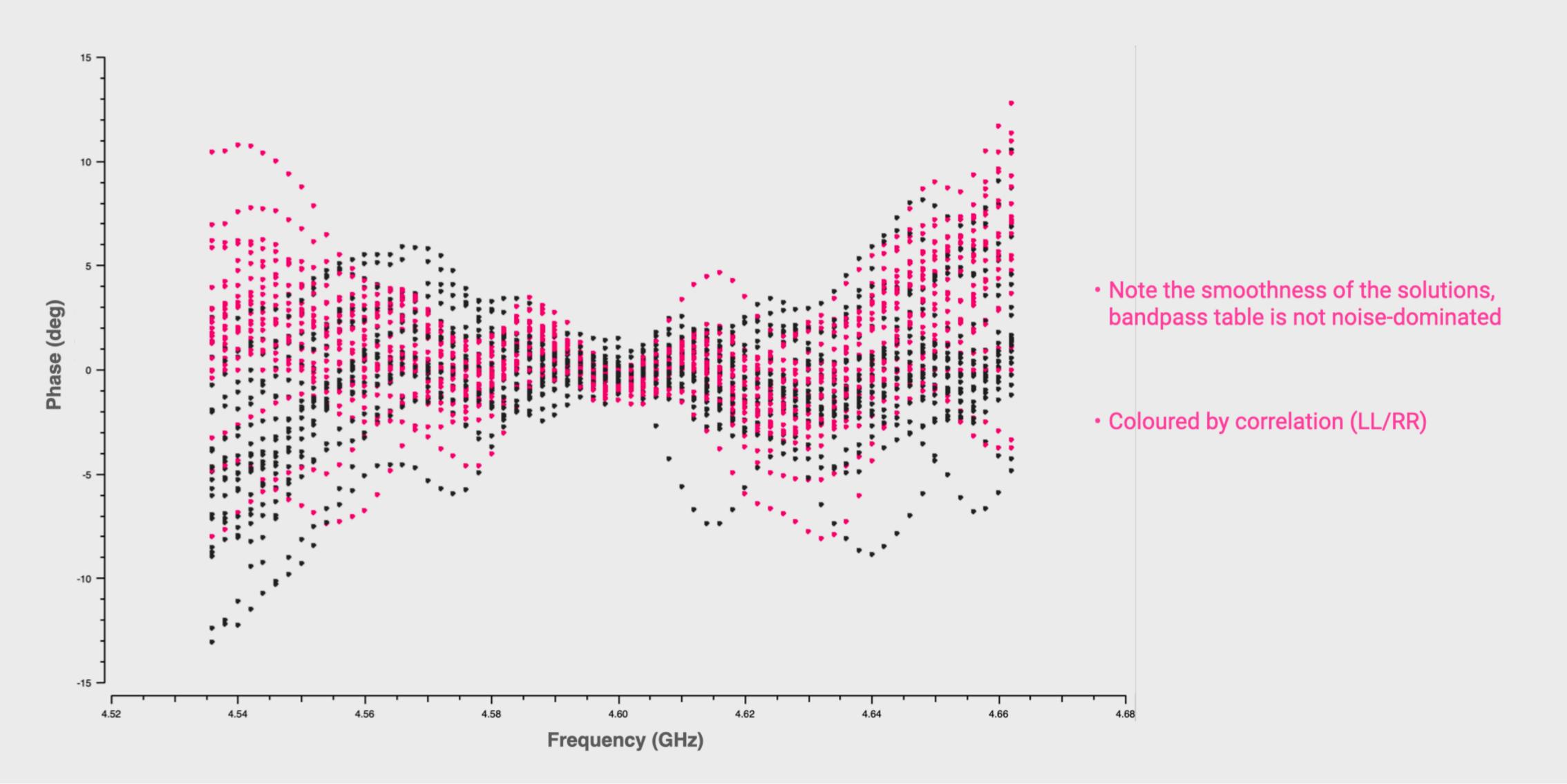
# Bandpass solutions (B): amplitude vs frequency



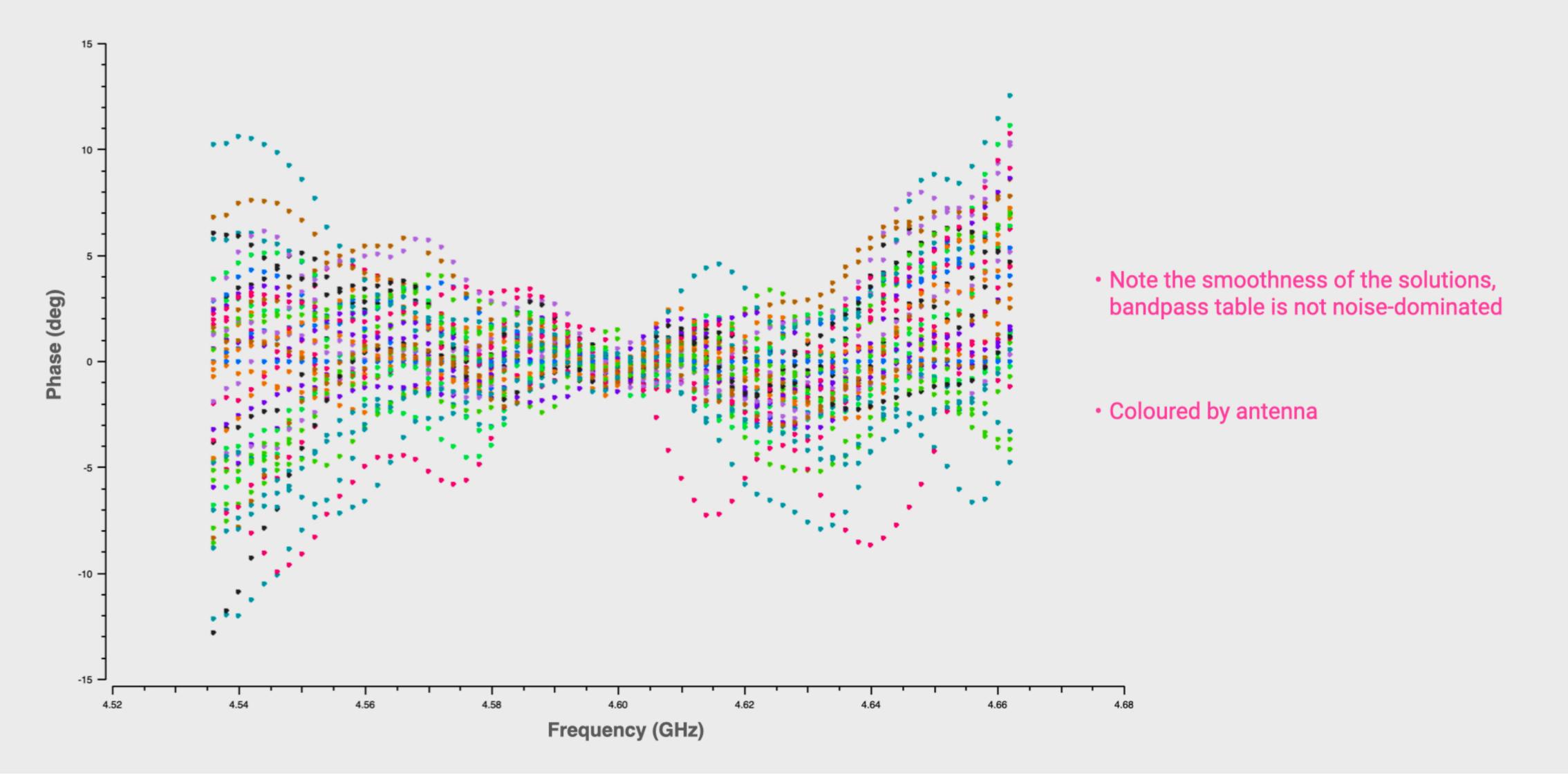
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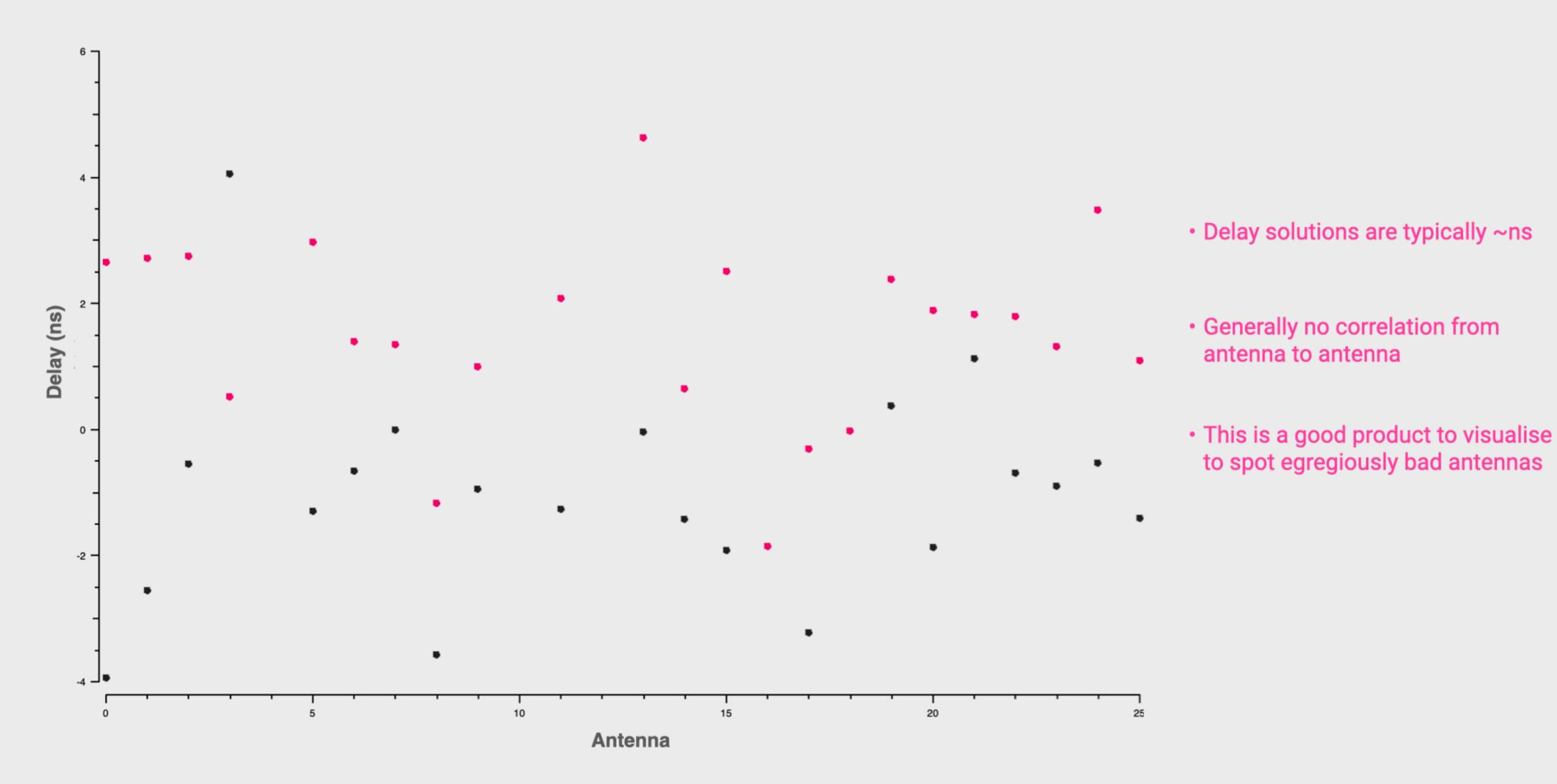
# Bandpass solutions (B): phase vs frequency



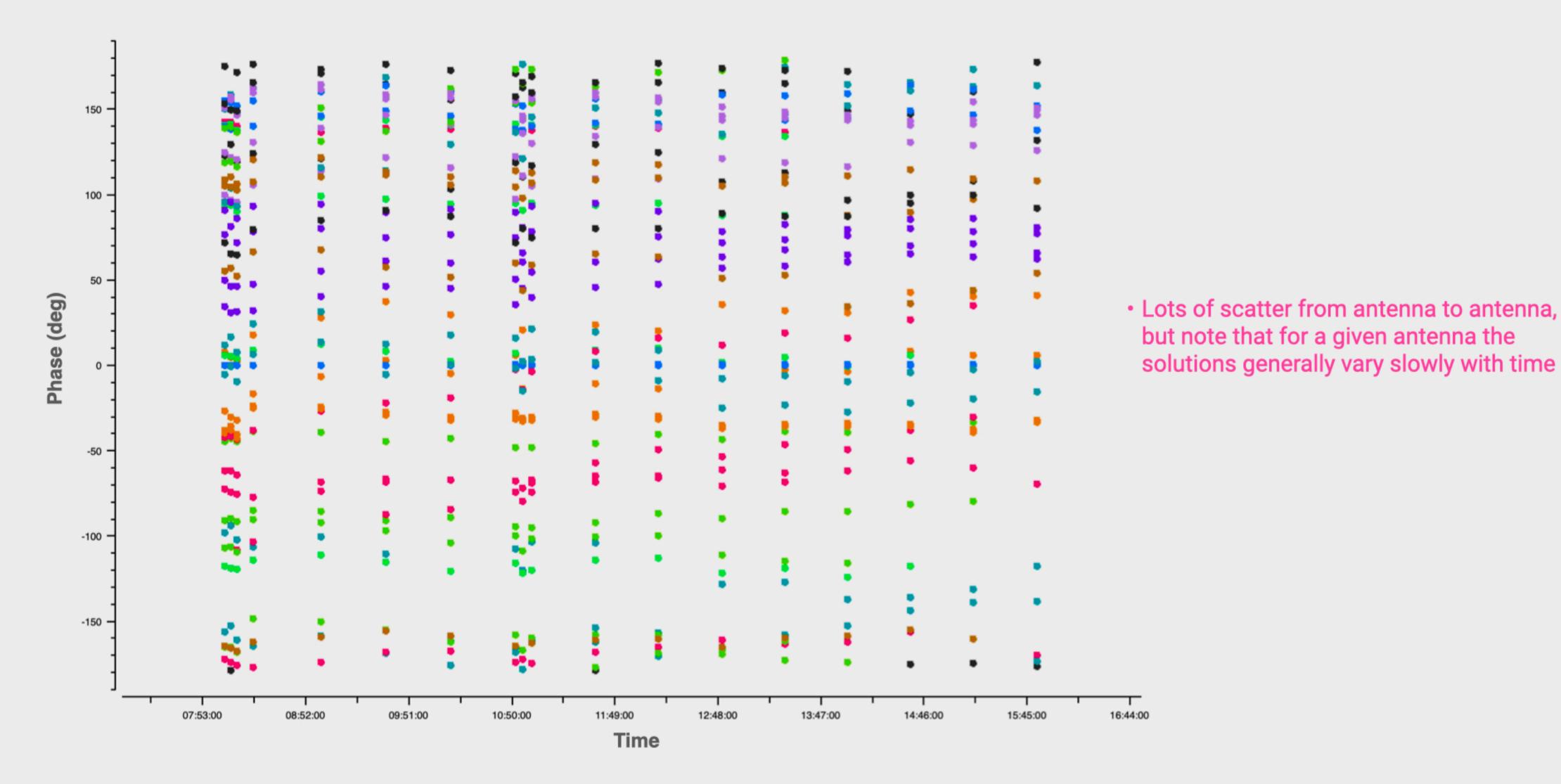
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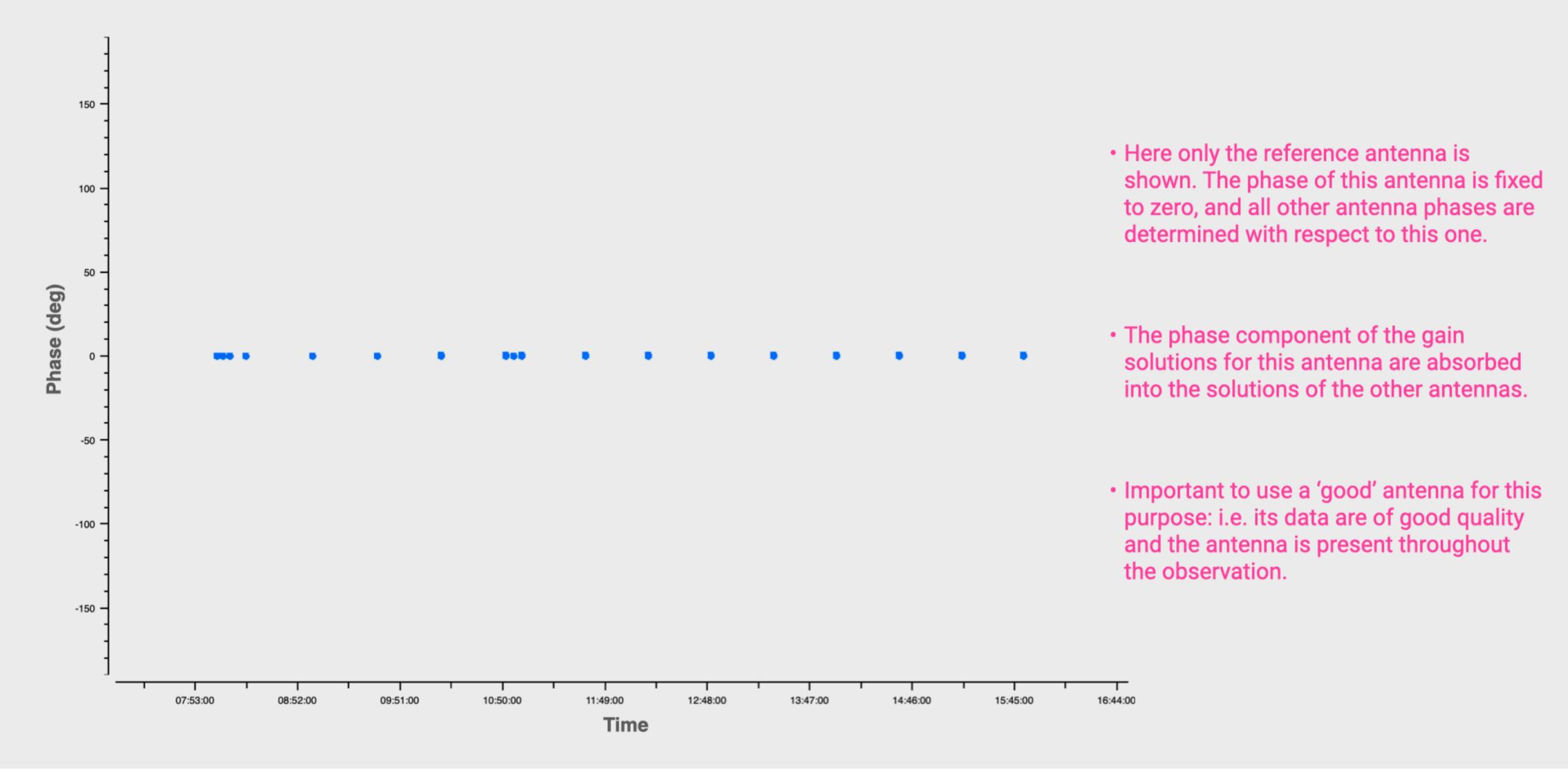
# Delay solutions (K): delay vs antenna

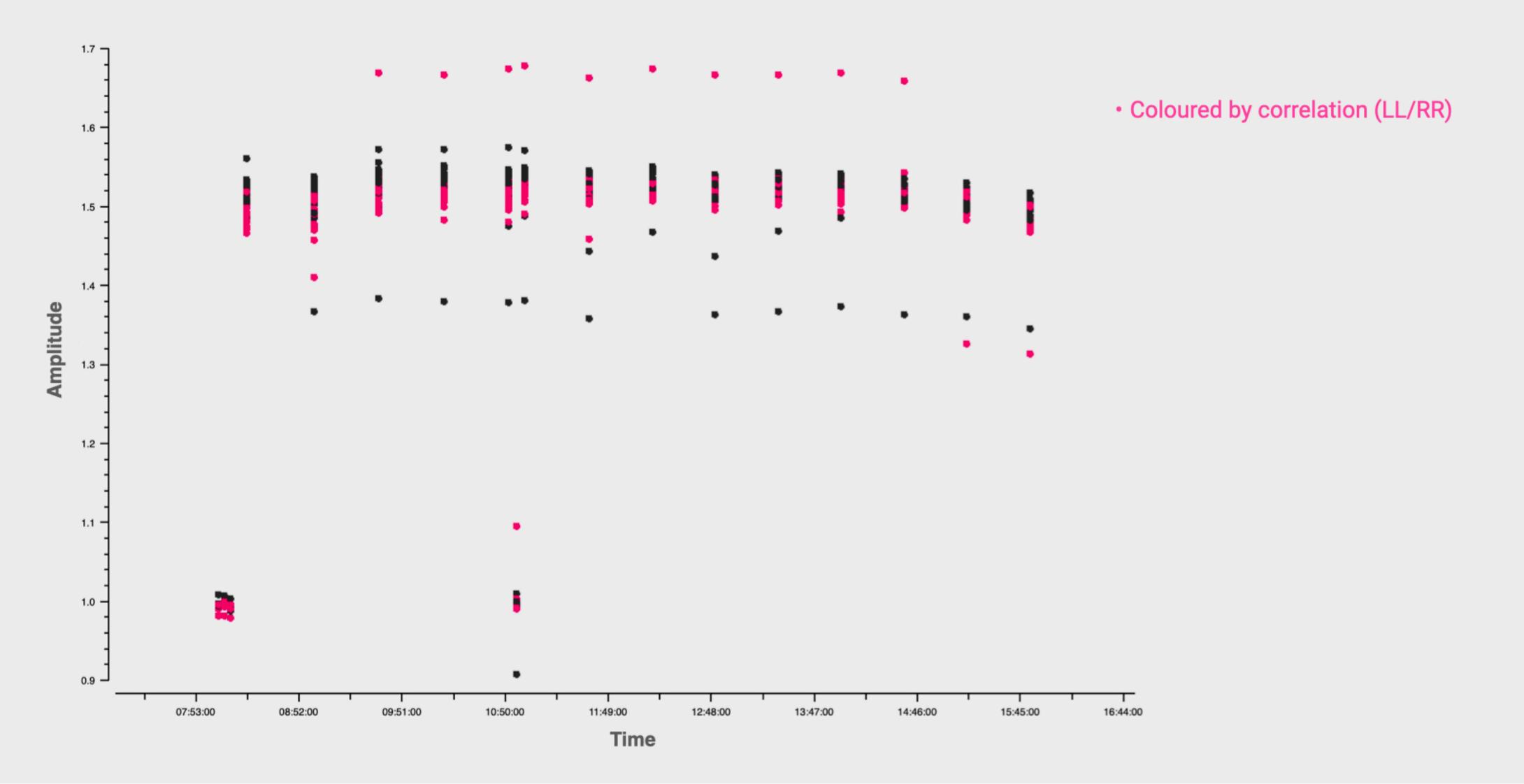


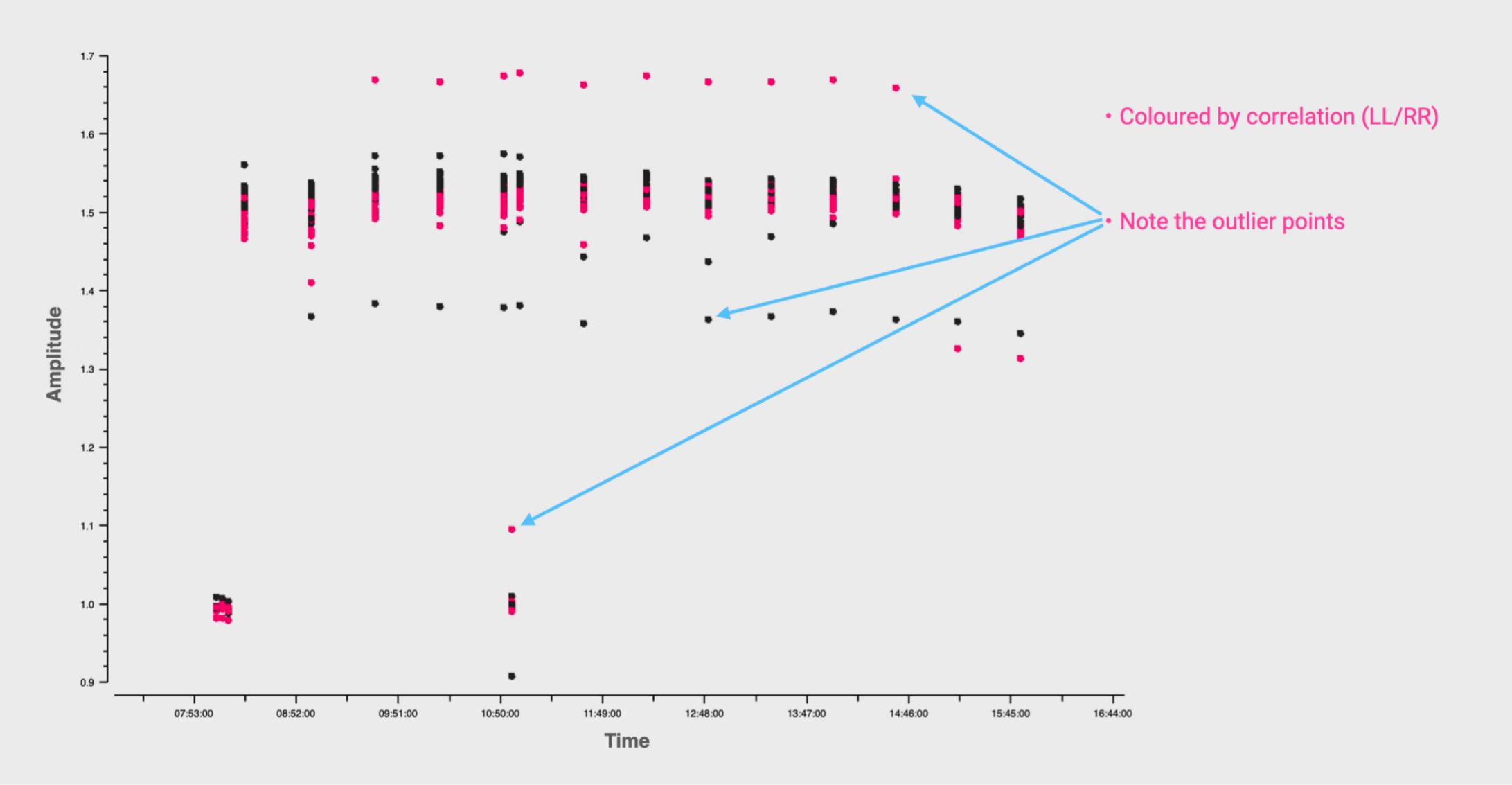
## Time dependent gains (G): phase vs time

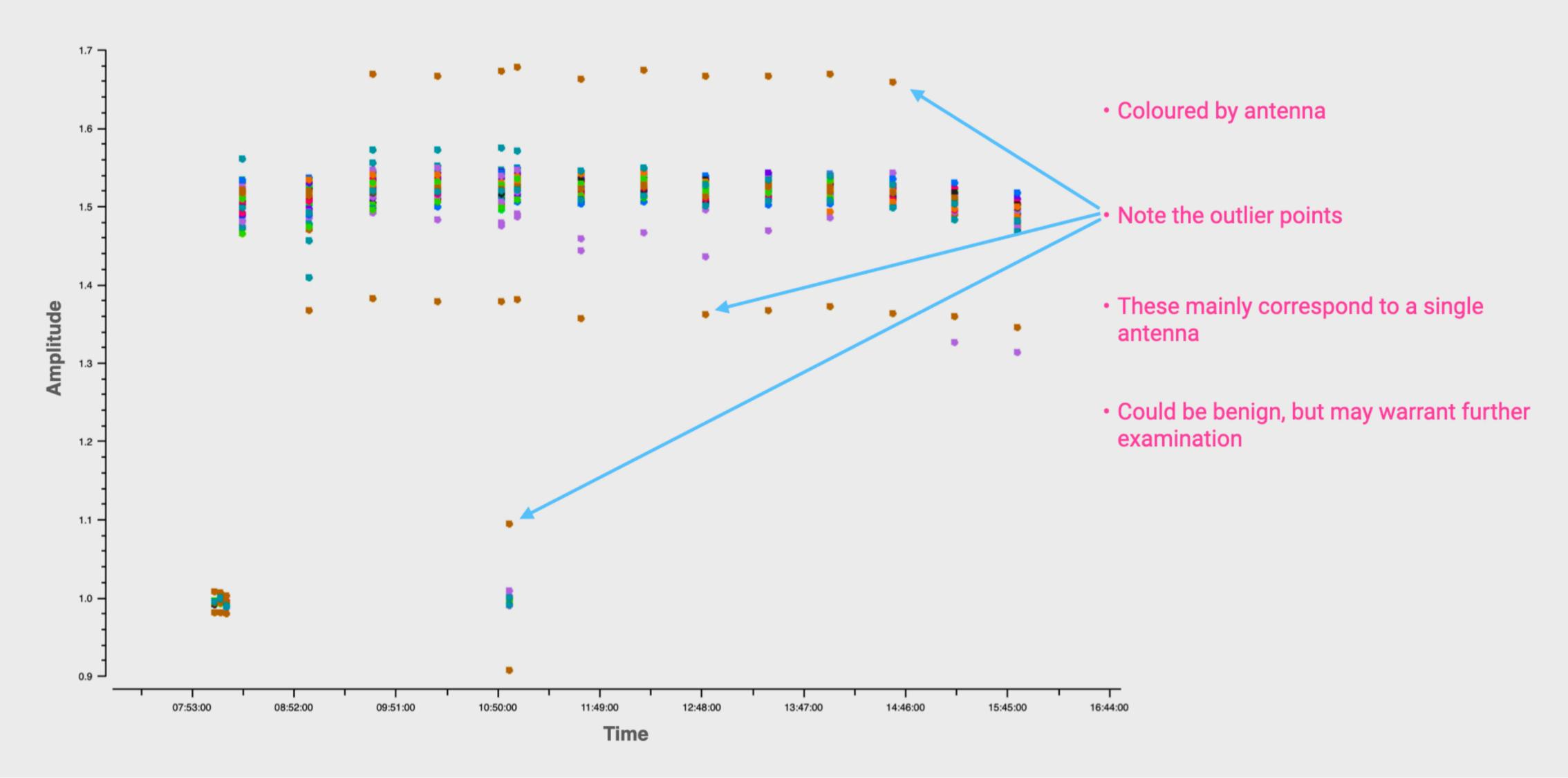


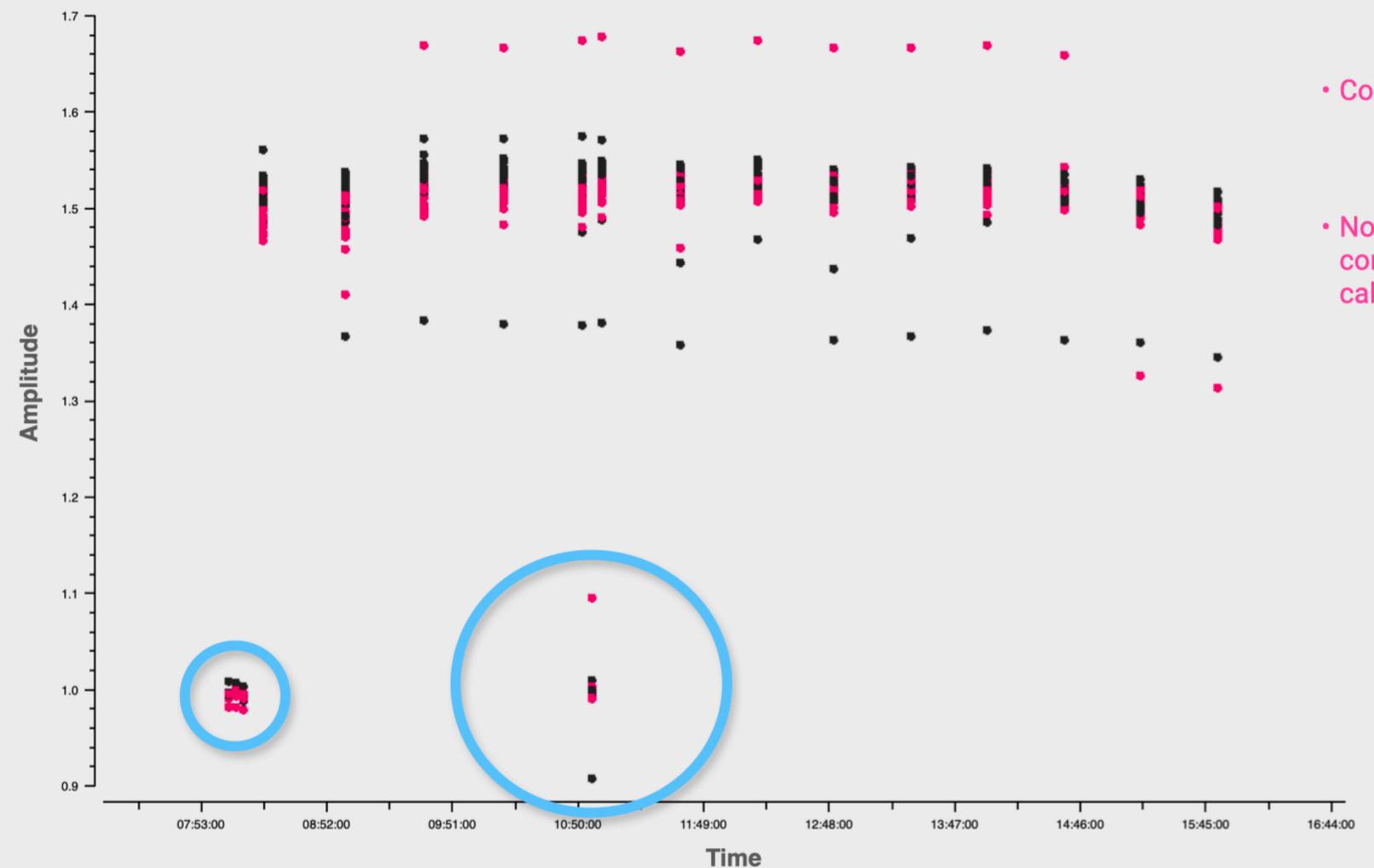
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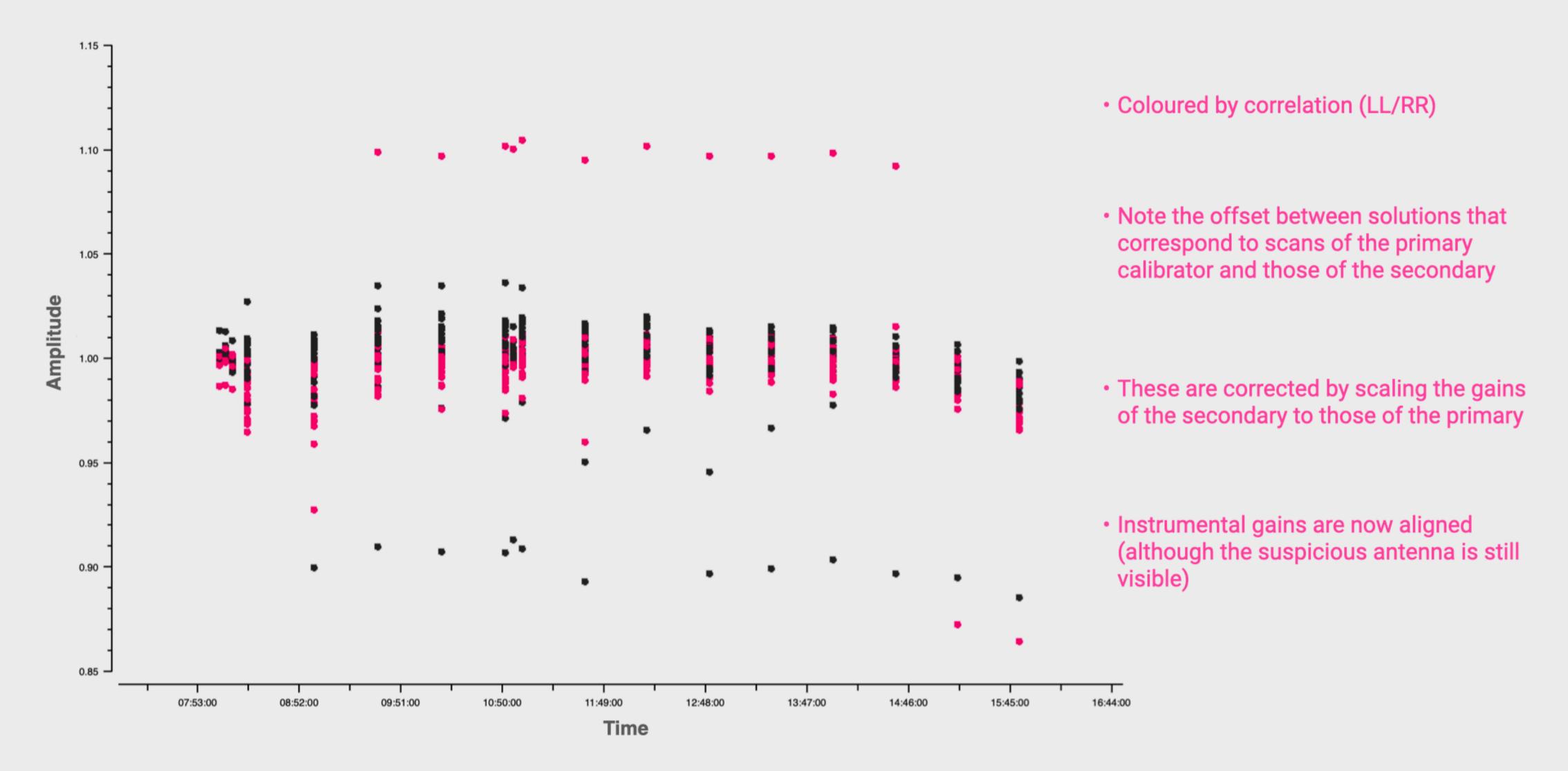




Coloured by correlation (LL/RR)

 Note the offset between solutions that correspond to scans of the primary calibrator and those of the secondary

### Fluxscale table: amplitude vs time



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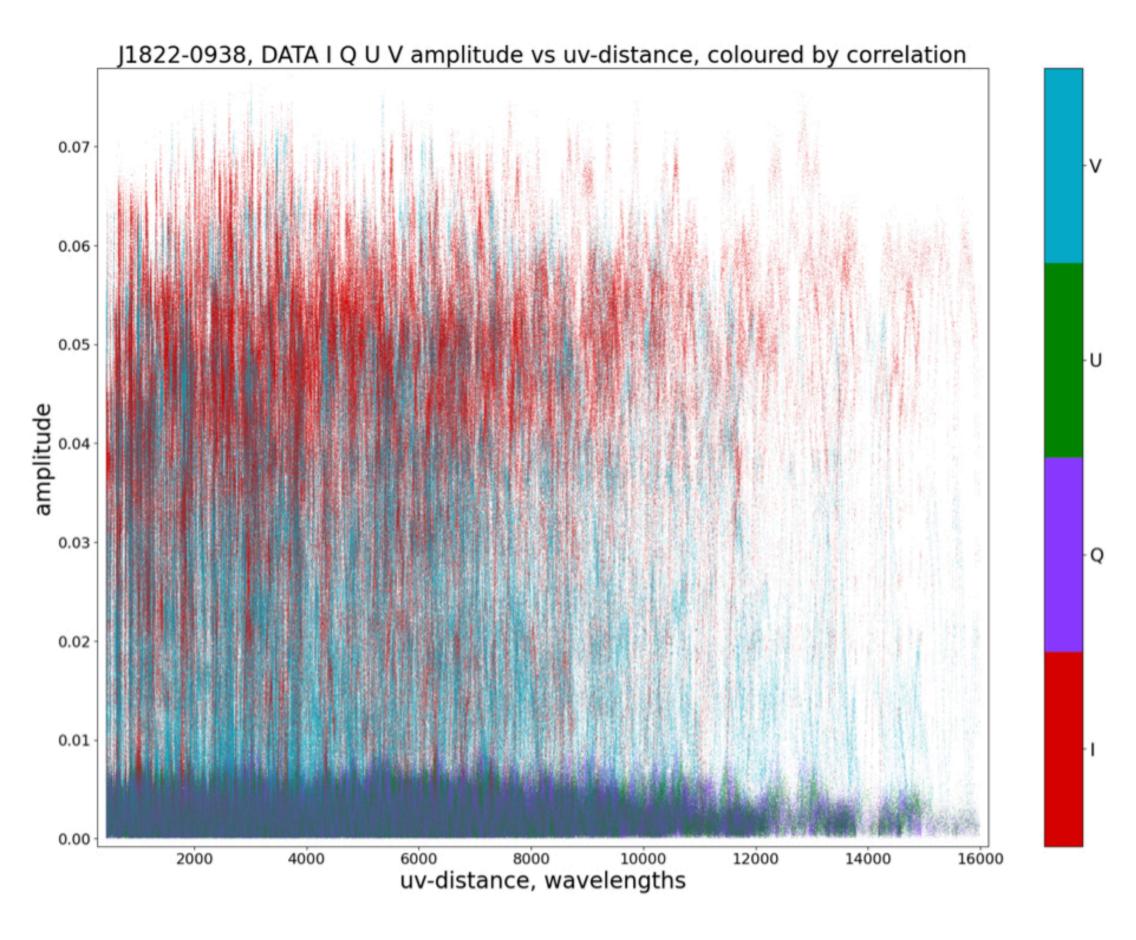
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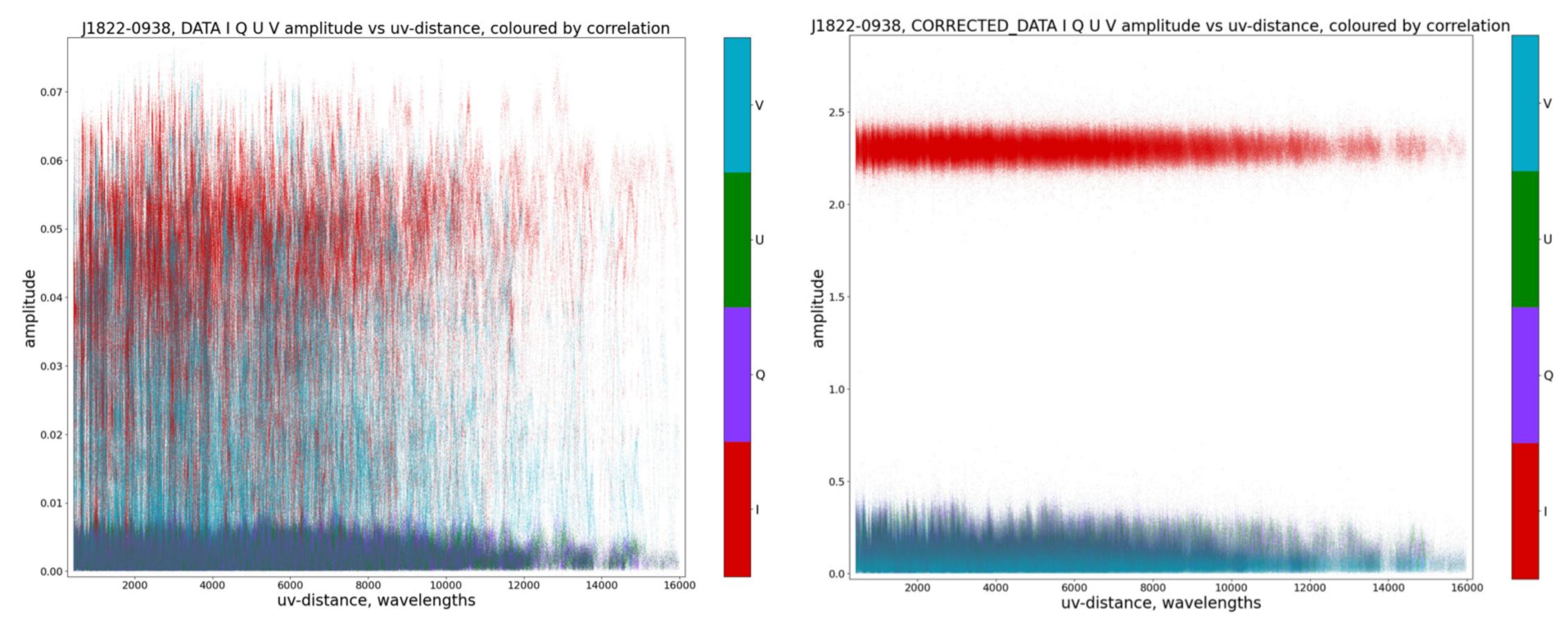
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  - Solutions must not be noise-dominated (acceptable SNR will depend on science goal)
  - Outlying points should be investigated (improve data flagging and iterate the calibration)

## Secondary calibrator (J1822-0938) visibilities: amplitude vs u,v distance



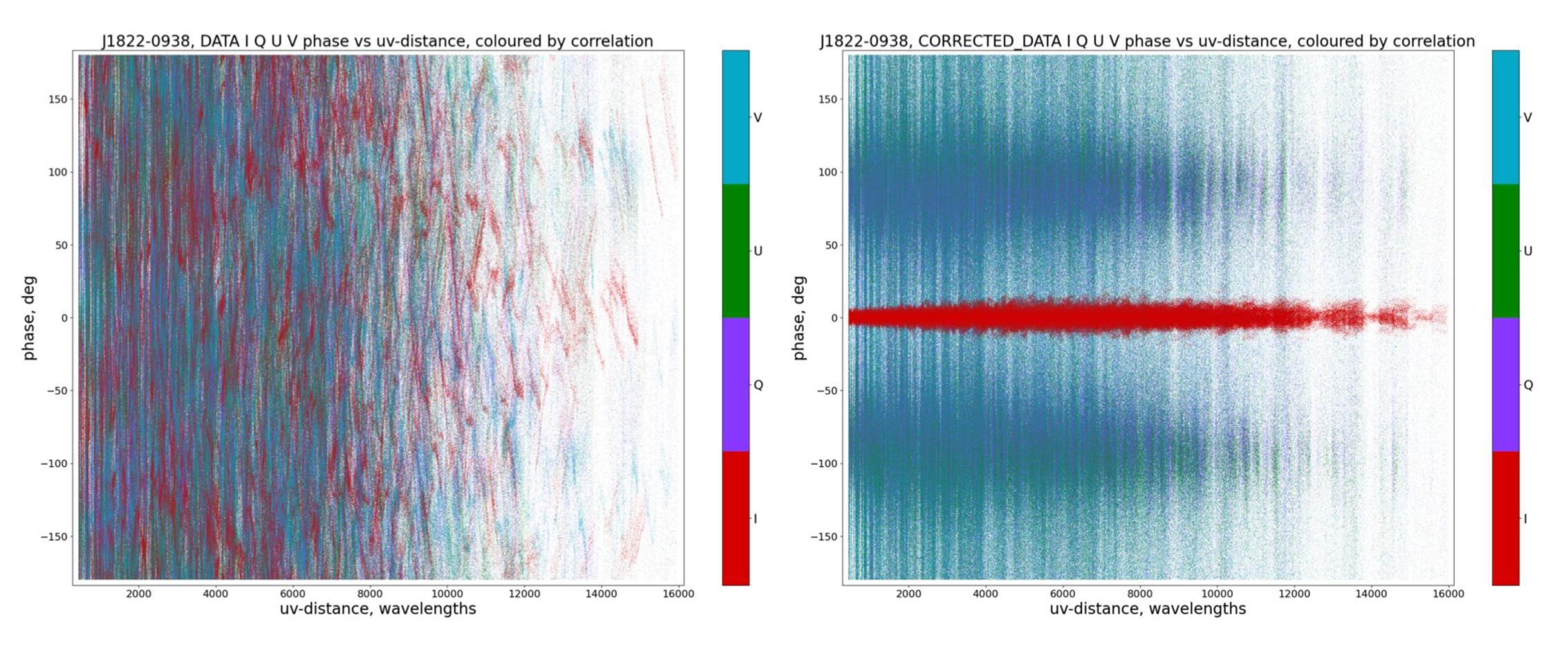
Uncalibrated

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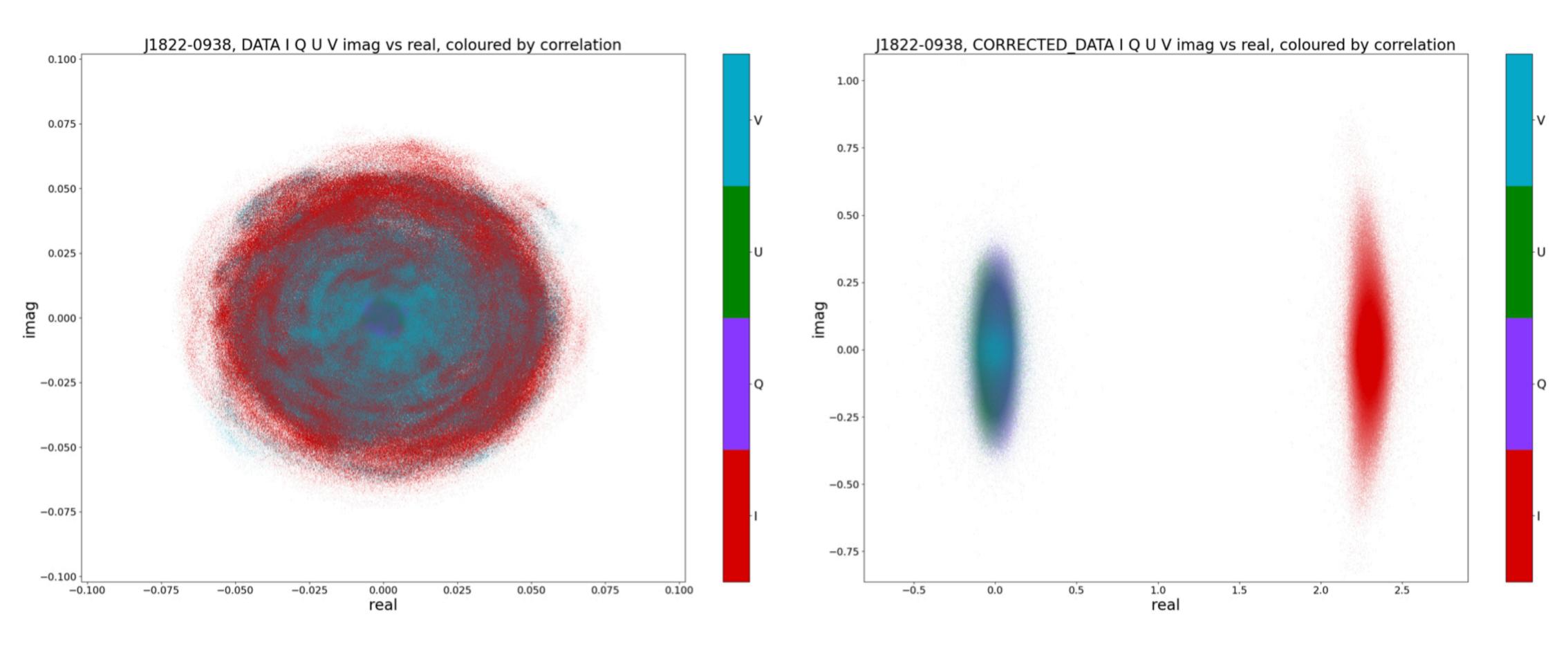
Uncalibrated Calibrated

## Secondary calibrator (J1822-0938) visibilities: phase vs u,v distance



Uncalibrated Calibrated

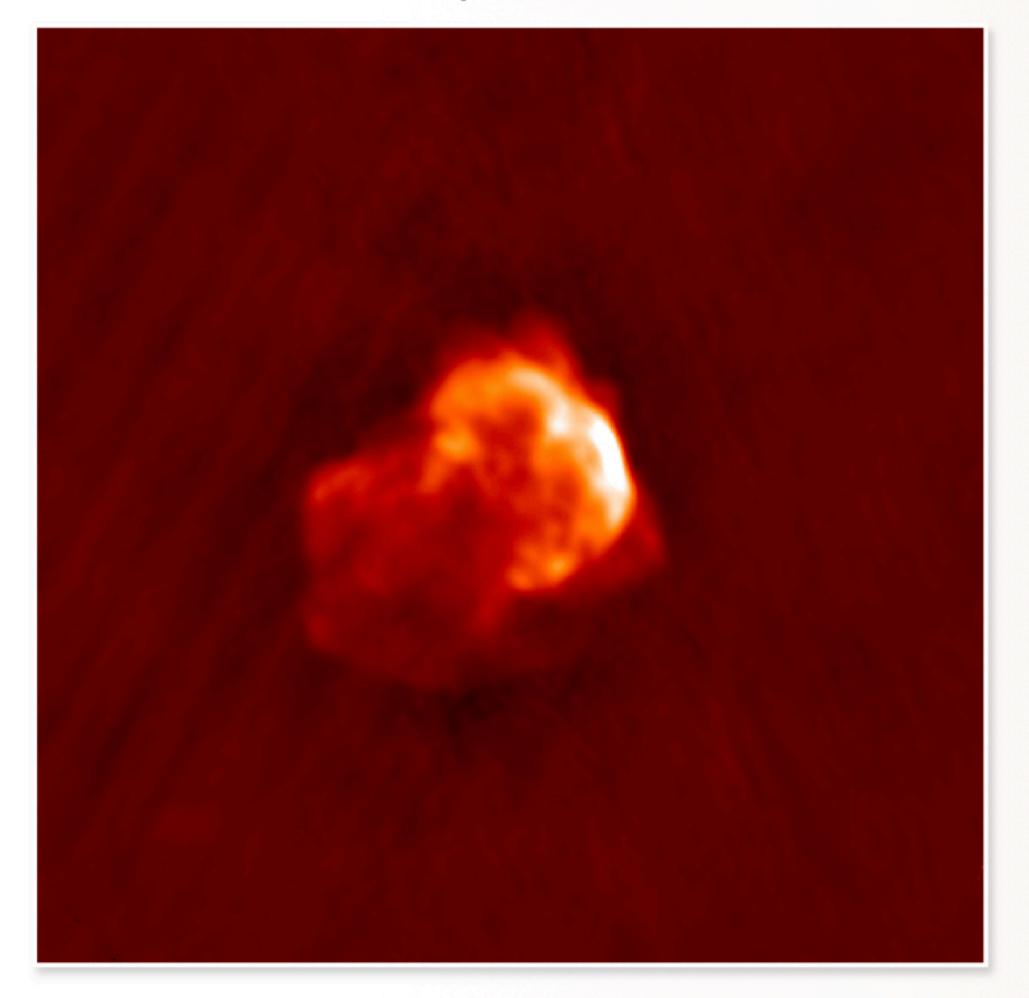
# Secondary calibrator (J1822-0938) visibilities: imaginary vs real



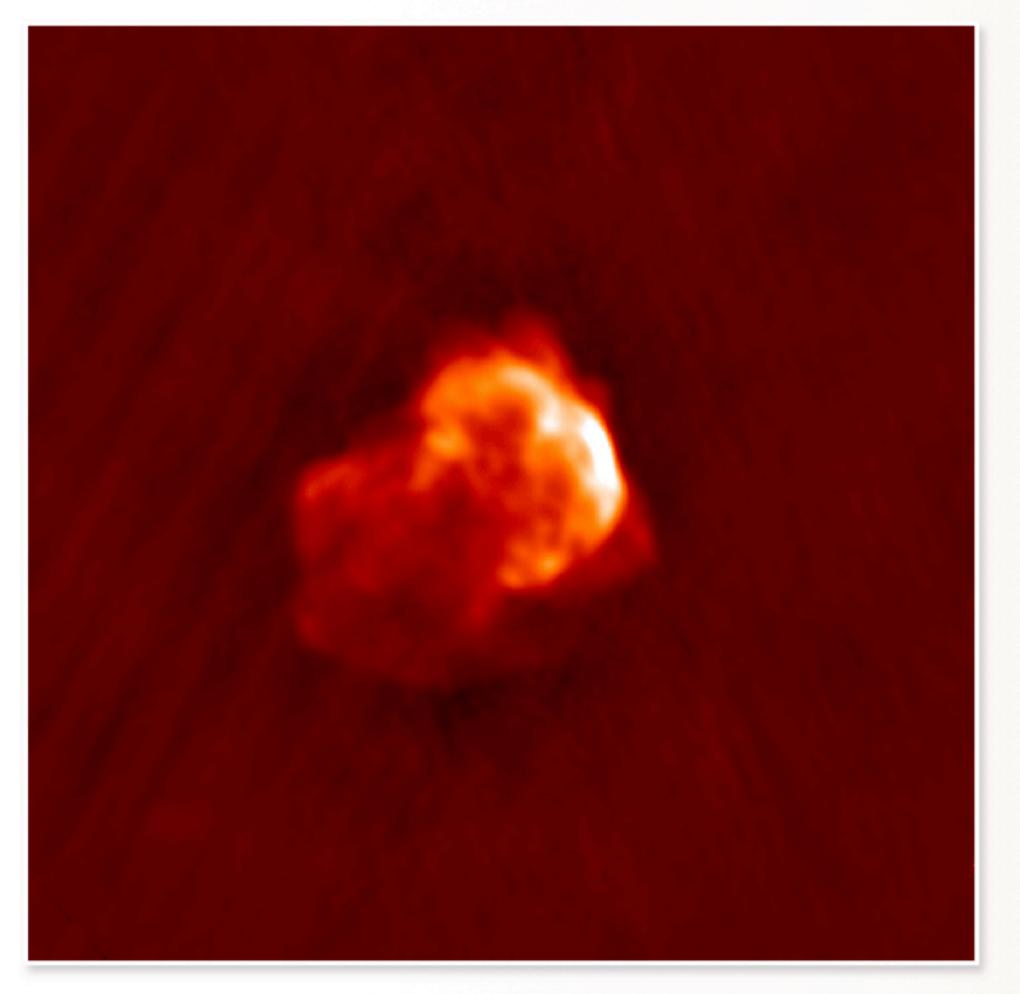
Uncalibrated Calibrated

The calibrated target, and to summarise...

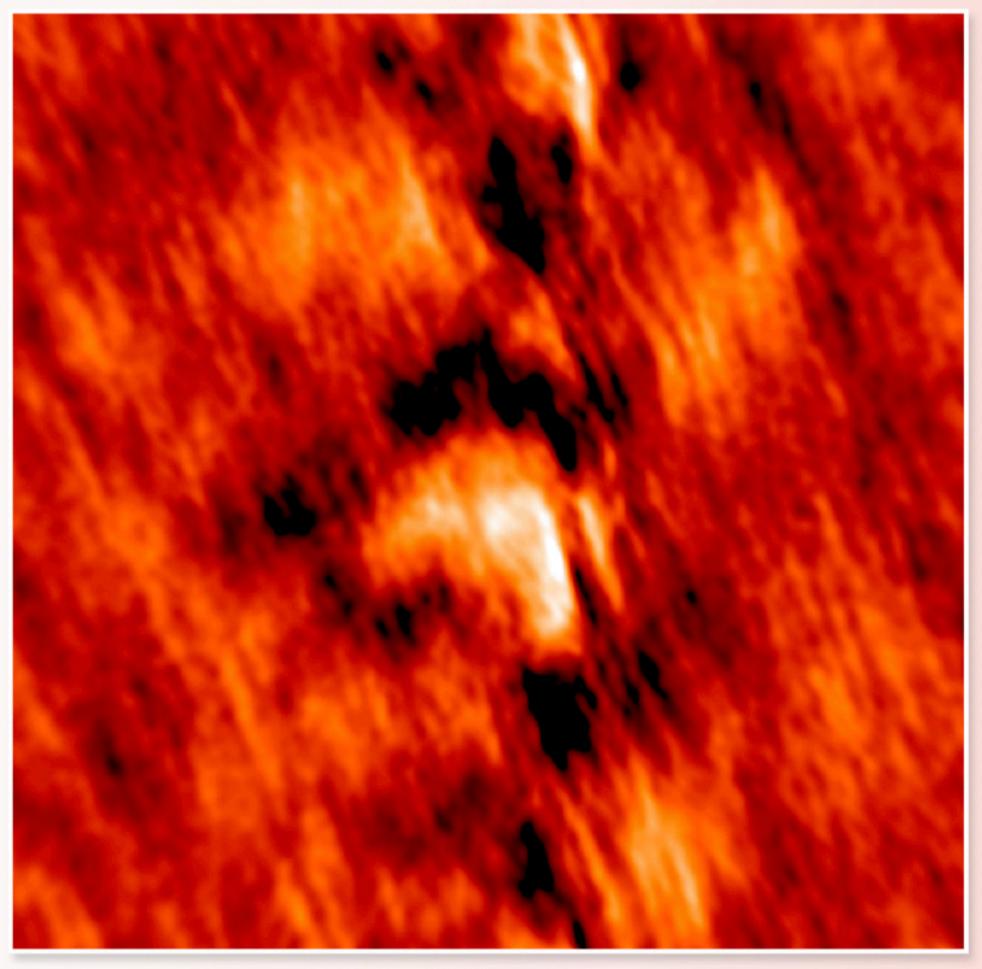
With proper calibration: beautiful supernova remnant



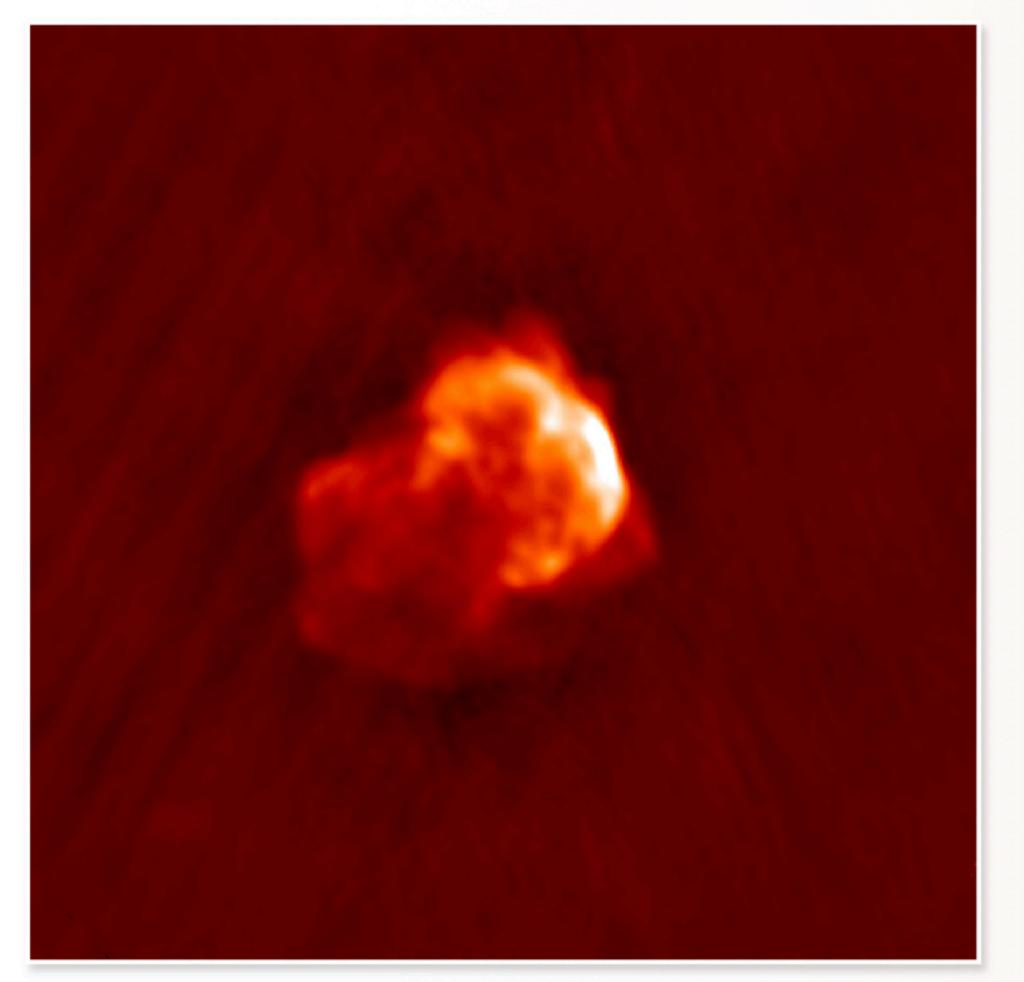
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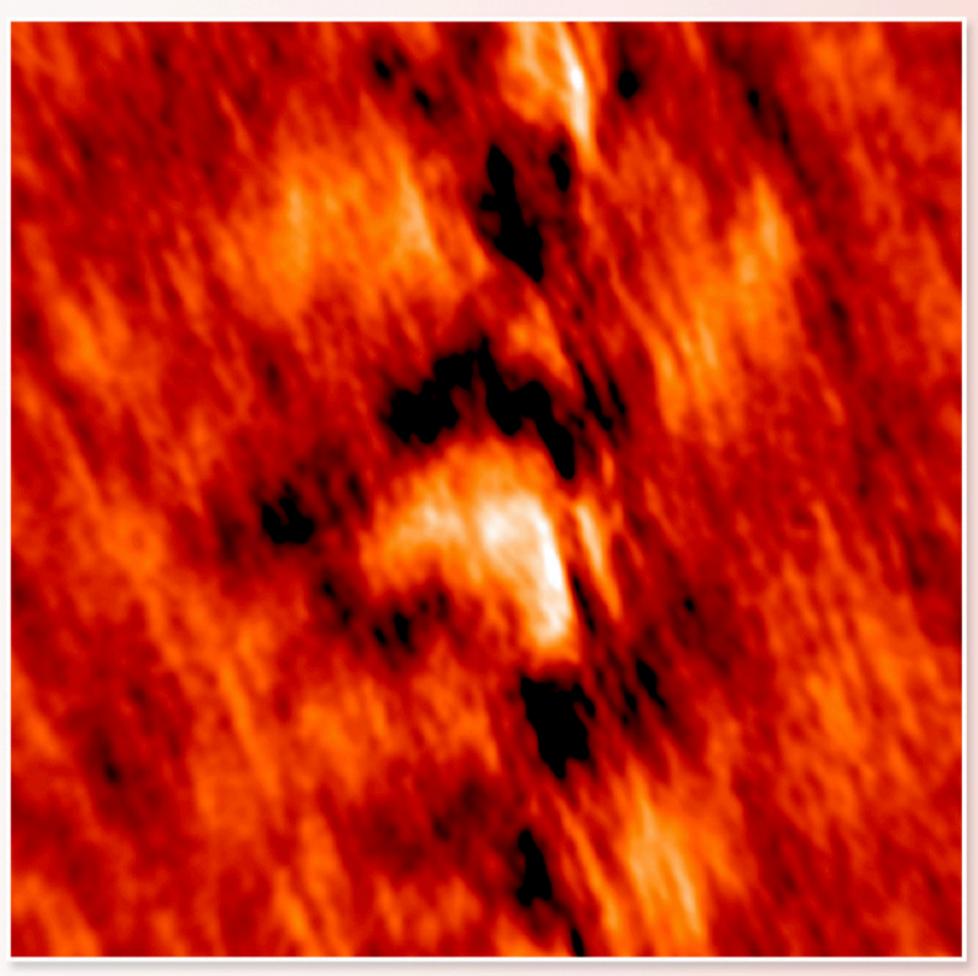
Without calibration: hot garbage



With proper calibration: beautiful supernova remnant



Without calibration: hot garbage



Calibration = good

"Why should I care? Isn't there a pipeline for this stuff?"

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  A working understanding of the telescope, and the calibration / imaging process is the only way to avoid this.
- If you do have to ask for help or advice (and we all do), then you will be better equipped to ask the right questions.

