



Calibration and Editing

George Moellenbrock

- Why calibration and editing?
- Formalism: Visibilities, signals, matrices
- Laundry List of Calibration Components
- Practical Calibration Planning
- Editing and RFI
- Calibration Sequence Examples
- Evaluating Calibration Performance
- Summary

Why Calibration and Editing?

- Synthesis radio telescopes, though well-designed, are not perfect (e.g., surface accuracy, receiver noise)
- Need to accommodate engineering (e.g., frequency conversion, digital electronics, etc.)
- Hardware or control software occasionally fails or behaves unpredictably
- Scheduling/observation errors sometimes occur (e.g., wrong source positions)
- Atmospheric conditions not ideal (not just bad weather)
- RFI

Determining instrumental properties (calibration)
is as important as
determining radio source properties

From Idealistic to Realistic

- Formally, we wish to obtain the visibility function, which we intend to invert to obtain an image of the sky:

$$V(u,v) = \int I_n(l,m) e^{-i2\pi(ul+vm)} dl dm$$

- In practice, we correlate the electric field (voltage) samples taken at pairs of telescopes (baselines $i-j$):

$$V_{ij}(t) = \left\langle x_i(t) e^{i f_i^{comp}} \cdot x_j^*(t) e^{-i f_j^{comp}} \right\rangle_t$$

- Single radio telescopes are devices for collecting the signal $x_i(t)$ and providing it to the correlator.

What signal is really collected?

- The net signal delivered by antenna i , $x_i(t)$, is a combination of the desired signal, $s_i(t,l,m)$, corrupted by a factor $J_i(t,l,m)$ and integrated over the sky, and noise, $n_i(t)$:

$$x_i(t) = \int J_i(t,l,m) s_i(t,l,m) dldm + n_i(t) \\ = s_i'(t) + n_i(t)$$

- $J_i(t,l,m)$ is the product of a host of effects which we must *calibrate*
- In some cases, effects contained in the $J_i(t,l,m)$ term corrupt the signal irreversibly and the resulting data must be *edited*
- $J_i(t,l,m)$ is a complex number
- $J_i(t,l,m)$ is *antenna-based*

Aside: Correlation of signals

- The correlation of two realistic signals from different antennas

$$\begin{aligned} \langle x_i \cdot x_j^* \rangle &= \langle (s_i' + n_i) \cdot (s_j' + n_j)^* \rangle \\ &= \langle s_i' \cdot s_j'^* \rangle + \langle s_i' \cdot n_j^* \rangle + \langle n_i \cdot s_j'^* \rangle + \langle n_i \cdot n_j^* \rangle \\ &= \langle s_i' \cdot s_j'^* \rangle \\ &= \left\langle \int J_i(t) s_i(t) dldm \cdot \int J_j^*(t) s_j^*(t) dldm \right\rangle \\ &= \left\langle \int J_i(t) J_j^*(t) s_i(t) s_j^*(t) dldm \right\rangle \end{aligned}$$

- Noise doesn't correlate—even if $n_i \gg s_i$, correlation isolates desired signals
- In integral, only $s_i(t,l,m)$, from the same directions correlate, so order of integration and signal product can be exchanged and terms re-ordered

- ...and the auto-correlation of a signal from a single antenna:

$$\langle x_i \cdot x_i^* \rangle = \langle s_i' \cdot s_i'^* \rangle + \langle n_i \cdot n_i^* \rangle$$

- Desired signal not isolated from noise (less useful!)

Formalism: Describe both polarizations via matrices

- Need two polarizations (p,q) to fully describe sampled EM wave front, where $p,q = R,L$ (circulars) or $p,q = X,Y$ (linears)
- Some components of J_i involve mixing of polarizations, so dual-polarization description desirable or even required

- So substitute:

$$s_i \rightarrow \vec{s}_i = \begin{pmatrix} s_i^p \\ s_i^q \end{pmatrix}, \quad J_i \rightarrow \vec{J}_i = \begin{pmatrix} J^{p \rightarrow p} & J^{q \rightarrow p} \\ J^{p \rightarrow q} & J^{q \rightarrow q} \end{pmatrix}$$

- The Jones matrix thus corrupts a signal as follows:

$$\begin{aligned} \vec{s}_i' &= \vec{J}_i \vec{s}_i \quad (i = m = 0, \text{ integral omitted}) \\ \begin{pmatrix} s_i'^p \\ s_i'^q \end{pmatrix} &= \begin{pmatrix} J^{p \rightarrow p} & J^{q \rightarrow p} \\ J^{p \rightarrow q} & J^{q \rightarrow q} \end{pmatrix} \begin{pmatrix} s_i^p \\ s_i^q \end{pmatrix} \\ &= \begin{pmatrix} J^{p \rightarrow p} s_i^p + J^{q \rightarrow p} s_i^q \\ J^{p \rightarrow q} s_i^p + J^{q \rightarrow q} s_i^q \end{pmatrix} \end{aligned}$$

Signal Correlation and Matrices

- Four correlations are possible from two polarizations. The *outer product* (a 'bookkeeping' product) forms them:

$$\left\langle \begin{pmatrix} s_i^p \\ s_i^q \end{pmatrix} \otimes \begin{pmatrix} s_j^p \\ s_j^q \end{pmatrix} \right\rangle = \left\langle \begin{pmatrix} s_i^p \\ s_i^q \end{pmatrix} \otimes \begin{pmatrix} s_j^p \\ s_j^q \end{pmatrix} \right\rangle = \begin{pmatrix} \langle s_i^p \cdot s_j^p \rangle & \langle s_i^p \cdot s_j^q \rangle \\ \langle s_i^q \cdot s_j^p \rangle & \langle s_i^q \cdot s_j^q \rangle \end{pmatrix}$$

- These four correlations (pp, pq, qp, qq) map to Stokes (I,Q,U,V) visibilities
- A very useful property of outer products:

$$(A \otimes B)(A' \otimes B') = (AA') \otimes (BB')$$

- (where A, B, A', B' are matrices and/or vectors of appropriate dimensions):

Signal Correlation and Matrices (cont)

- The outer product for the Jones matrix:

$$\begin{aligned} \tilde{J}_i \otimes \tilde{J}_j^* &= \begin{pmatrix} J_i^{p \rightarrow p} & J_i^{q \rightarrow p} \\ J_i^{p \rightarrow q} & J_i^{q \rightarrow q} \end{pmatrix} \otimes \begin{pmatrix} J_j^{p \rightarrow p} & J_j^{q \rightarrow p} \\ J_j^{p \rightarrow q} & J_j^{q \rightarrow q} \end{pmatrix}^* \\ &= \begin{pmatrix} J_i^{p \rightarrow p} J_j^{p \rightarrow p} & J_i^{p \rightarrow p} J_j^{q \rightarrow p} & J_i^{q \rightarrow p} J_j^{p \rightarrow p} & J_i^{q \rightarrow p} J_j^{q \rightarrow p} \\ J_i^{p \rightarrow p} J_j^{p \rightarrow q} & J_i^{p \rightarrow p} J_j^{q \rightarrow q} & J_i^{q \rightarrow p} J_j^{p \rightarrow q} & J_i^{q \rightarrow p} J_j^{q \rightarrow q} \\ J_i^{p \rightarrow q} J_j^{p \rightarrow p} & J_i^{p \rightarrow q} J_j^{q \rightarrow p} & J_i^{q \rightarrow q} J_j^{p \rightarrow p} & J_i^{q \rightarrow q} J_j^{q \rightarrow p} \\ J_i^{p \rightarrow q} J_j^{p \rightarrow q} & J_i^{p \rightarrow q} J_j^{q \rightarrow q} & J_i^{q \rightarrow q} J_j^{p \rightarrow q} & J_i^{q \rightarrow q} J_j^{q \rightarrow q} \end{pmatrix} = \tilde{J}_{ij} \end{aligned}$$

- \tilde{J}_{ij} is a 4x4 Mueller matrix
- Antenna and array design thankfully driven by minimizing off-diagonal terms

Signal Correlation and Matrices (cont)

- And finally, for fun, the correlation of corrupted signals:

$$\begin{aligned} \tilde{J}_i \tilde{J}_j^* &= \left(\tilde{J}_i \otimes \tilde{J}_j^* \right) \left(\tilde{J}_i \otimes \tilde{J}_j^* \right)^* \\ &= \begin{pmatrix} J_i^{p \rightarrow p} J_j^{p \rightarrow p} & J_i^{p \rightarrow p} J_j^{q \rightarrow p} & J_i^{q \rightarrow p} J_j^{p \rightarrow p} & J_i^{q \rightarrow p} J_j^{q \rightarrow p} \\ J_i^{p \rightarrow p} J_j^{p \rightarrow q} & J_i^{p \rightarrow p} J_j^{q \rightarrow q} & J_i^{q \rightarrow p} J_j^{p \rightarrow q} & J_i^{q \rightarrow p} J_j^{q \rightarrow q} \\ J_i^{p \rightarrow q} J_j^{p \rightarrow p} & J_i^{p \rightarrow q} J_j^{q \rightarrow p} & J_i^{q \rightarrow q} J_j^{p \rightarrow p} & J_i^{q \rightarrow q} J_j^{q \rightarrow p} \\ J_i^{p \rightarrow q} J_j^{p \rightarrow q} & J_i^{p \rightarrow q} J_j^{q \rightarrow q} & J_i^{q \rightarrow q} J_j^{p \rightarrow q} & J_i^{q \rightarrow q} J_j^{q \rightarrow q} \end{pmatrix} \begin{pmatrix} s_i^p \cdot s_j^p \\ s_i^p \cdot s_j^q \\ s_i^q \cdot s_j^p \\ s_i^q \cdot s_j^q \end{pmatrix} \\ &= \begin{pmatrix} J_i^{p \rightarrow p} J_j^{p \rightarrow p} s_i^p \cdot s_j^p + J_i^{p \rightarrow p} J_j^{q \rightarrow p} s_i^p \cdot s_j^q + J_i^{q \rightarrow p} J_j^{p \rightarrow p} s_i^q \cdot s_j^p + J_i^{q \rightarrow p} J_j^{q \rightarrow p} s_i^q \cdot s_j^q \\ J_i^{p \rightarrow p} J_j^{p \rightarrow q} s_i^p \cdot s_j^p + J_i^{p \rightarrow p} J_j^{q \rightarrow q} s_i^p \cdot s_j^q + J_i^{q \rightarrow p} J_j^{p \rightarrow q} s_i^q \cdot s_j^p + J_i^{q \rightarrow p} J_j^{q \rightarrow q} s_i^q \cdot s_j^q \\ J_i^{p \rightarrow q} J_j^{p \rightarrow p} s_i^p \cdot s_j^p + J_i^{p \rightarrow q} J_j^{q \rightarrow p} s_i^p \cdot s_j^q + J_i^{q \rightarrow q} J_j^{p \rightarrow p} s_i^q \cdot s_j^p + J_i^{q \rightarrow q} J_j^{q \rightarrow p} s_i^q \cdot s_j^q \\ J_i^{p \rightarrow q} J_j^{p \rightarrow q} s_i^p \cdot s_j^p + J_i^{p \rightarrow q} J_j^{q \rightarrow q} s_i^p \cdot s_j^q + J_i^{q \rightarrow q} J_j^{p \rightarrow q} s_i^q \cdot s_j^p + J_i^{q \rightarrow q} J_j^{q \rightarrow q} s_i^q \cdot s_j^q \end{pmatrix} \end{aligned}$$

- UGLY, but let's think about individual calibration components in the signal domain, where the matrices are a factor of 2 less complicated.....

Calibration Components

- \tilde{J}_i contains many components:

- F = ionospheric Faraday rotation
- T = tropospheric effects
- P = parallactic angle
- E = antenna voltage pattern
- D = polarization leakage
- G = electronic gain
- B = bandpass response

$$\tilde{J}_i = \tilde{B}_i \tilde{G}_i \tilde{D}_i \tilde{E}_i \tilde{P}_i \tilde{T}_i \tilde{F}_i$$

- Order of terms follows signal path
- Each term has matrix form of \tilde{J}_i with terms embodying its particular algebra (on- vs. off-diagonal terms, etc.)
- The full matrix equation (especially after correlation!) is daunting, but usually only need to consider the terms individually or in pairs, and rarely in open form (matrix formulation = shorthand)

Ionospheric Faraday Rotation, F

- The ionosphere is *birefringent*; one hand of circular polarization is delayed w.r.t. the other, introducing a phase shift:

$$\Delta F = 0.15 \int B n_{ds} \text{ deg}$$

I in cm, n_{ds} in 10^{14} cm^{-2} , B in G

- Rotates the linear polarization position angle
- More important at longer wavelengths:

$$TEC = \int n_{ds} \sim 10^{14} \text{ cm}^{-2}; B_i \sim 1 \text{ G}; I = 20 \text{ km} \rightarrow \Delta F \sim 60^\circ$$

- More important at solar maximum and at sunrise/sunset, when ionosphere is most active
- Beware of 'patchiness' and other variability (e.g., with elevation changes)
- Namir's lecture: "Long Wavelength Interferometry" (next Tuesday)

$$\tilde{F} = \begin{pmatrix} e^{i\Delta F} & 0 \\ 0 & e^{-i\Delta F} \end{pmatrix}; \tilde{F}^* = \begin{pmatrix} \cos \Delta F & -\sin \Delta F \\ \sin \Delta F & \cos \Delta F \end{pmatrix}$$

Tropospheric Effects, T

- The troposphere causes polarization-independent amplitude and phase effects due to emission/opacity and refraction, respectively

- Typically 2-3m excess path length at zenith compared to vacuum
- Most important at $\nu > 15$ GHz where water vapor absorbs/emits
- More important nearer horizon where tropospheric path length greater
- Clouds, weather = variability in phase and opacity; may vary across array
- Water vapor radiometry? Phase transfer from low to high frequencies?
- » Claire's lecture: "mm-Wave Interferometry" (next Monday)

$$\tilde{T}^{pq} = \begin{pmatrix} t & 0 \\ 0 & t \end{pmatrix} = t \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix}$$

Parallactic Angle, P

- Orientation of sky in telescope's field of view

- Constant for equatorial telescopes
- Varies for alt-az-mounted telescopes:

$$c(t) = \arctan\left(\frac{\cos(l)\sin(h(t))}{\sin(l)\cos(d) - \cos(l)\sin(d)\cos(h(t))}\right)$$

l = latitude, $h(t)$ = hour angle, d = declination

- Rotates the position angle of linearly polarized radiation (c.f. F)
- Analytically known, and its variation provides leverage for determining polarization-dependent effects

$$\tilde{P}^{RL} = \begin{pmatrix} e^{ic} & 0 \\ 0 & e^{-ic} \end{pmatrix}; \quad \tilde{P}^{XY} = \begin{pmatrix} \cos c & -\sin c \\ \sin c & \cos c \end{pmatrix}$$

Antenna Voltage Pattern, E

- Antennas of all designs have direction-dependent gain

- Important when region of interest on sky comparable to or larger than λ/D
 - » Kumar's lecture: "Wide Field Imaging I" (next Monday)
 - » Debra's lecture: "Wide Field Imaging II" (next Monday)
- Important at lower frequencies where radio source surface density is greater and wide-field imaging techniques required
- Beam squint: B and E^p not parallel, yielding spurious polarization
- For convenience, direction dependence of polarization leakage (D) may be included in E (off-diagonal terms then non-zero)

$$E^{pq} = \begin{pmatrix} e^p(l, m) & 0 \\ 0 & e^q(l, m) \end{pmatrix}$$

Polarization Leakage, D

- Polarizer is not ideal, so orthogonal polarizations not perfectly isolated

- Well-designed feeds have $d \sim$ a few percent or less
- A geometric property of the feed design, so frequency dependent
- For R,L systems, total-intensity imaging affected as $\sim dQ, dU$, so only important at high dynamic range (because $Q, U \sim d$, typically)
- For R,L systems, linear polarization imaging affected as $\sim dI$, so almost always important
 - » Greg's lecture: "Polarization in Interferometry" (today!)

$$\tilde{D}^{pq} = \begin{pmatrix} 1 & d^p \\ d^q & 1 \end{pmatrix}$$

Electronic Gain, G

- Catch-all for most amplitude and phase effects introduced by antenna electronics (amplifiers, mixers, quantizers, digitizers)

- Most commonly treated calibration component
- Dominates other effects for standard observations
- Includes scaling from engineering to radioastronomy units (Jy)
- Often includes ionospheric and tropospheric effects which are typically difficult to separate unto themselves
- Excludes frequency dependent effects (see B)

$$G^{pq} = \begin{pmatrix} g^p & 0 \\ 0 & g^q \end{pmatrix}$$

Bandpass Response, B

- G -like component describing frequency-dependence of antenna electronics, etc.

- Filters used to select frequency passband not square
- Optical and electronic reflections introduce ripples across band
- Typically (but not necessarily) normalized

$$B^{pq} = \begin{pmatrix} b^p(\mathbf{n}) & 0 \\ 0 & b^q(\mathbf{n}) \end{pmatrix}$$

More-sophisticated effects

- Errors in geometric/clock models in correlator cause poor phase compensation

- Routine problem in VLBI solved by fringe-fitting: parameterization of G to include phase terms which are linear in time and frequency
 - » Craig's lecture: "VLBI" (Thursday)

- Baseline-based errors do not decompose into antenna-based components

- Most digital correlators designed to limit such effects to well-understood and uniform scaling laws (absorbed in G)
- Additional errors can result from averaging in time and frequency over variation in antenna-based effects and visibilities (practical instruments are finite!)
- Correlated noise (e.g., RFI)
- Virtually indistinguishable from source structure effects
- Geodetic observers consider determination of radio source structure—a baseline-based effect—as a required *calibration* if antenna positions are to be determined accurately

Putting it all back together

- In the correlation of signals, like terms from the different antennas are conveniently grouped:

$$\begin{aligned} \tilde{J}_i \otimes \tilde{J}_j &= \tilde{B}_i \tilde{G}_i \tilde{D}_i \tilde{E}_i \tilde{P} \tilde{T}_i \tilde{F}_i \otimes \tilde{B}_j \tilde{G}_j \tilde{D}_j \tilde{E}_j \tilde{P} \tilde{T}_j \tilde{F}_j \\ &= (\tilde{B}_i \otimes \tilde{B}_j) (\tilde{G}_i \otimes \tilde{G}_j) (\tilde{D}_i \otimes \tilde{D}_j) (\tilde{E}_i \otimes \tilde{E}_j) (\tilde{P}_i \otimes \tilde{P}_j) (\tilde{T}_i \otimes \tilde{T}_j) (\tilde{F}_i \otimes \tilde{F}_j) \\ &= \tilde{B}_{ij} \tilde{G}_{ij} \tilde{D}_{ij} \tilde{E}_{ij} \tilde{P}_{ij} \tilde{T}_{ij} \tilde{F}_{ij} \end{aligned}$$

- The total *Measurement Equation* has the form:

$$\tilde{V}_{ij} = \tilde{M}_{ij} \int \tilde{B}_{ij} \tilde{G}_{ij} \tilde{D}_{ij} \tilde{E}_{ij} \tilde{P}_{ij} \tilde{T}_{ij} \tilde{F}_{ij} \tilde{S}_n(l, m) e^{-i2\pi(u_{ij}l + v_{ij}m)} dl dm + \tilde{A}_{ij}$$

- \tilde{S} maps the Stokes vector, I , to the polarization basis of the instrument
- \tilde{M}_{ij} and \tilde{A}_{ij} are multiplicative and additive baseline-based errors, respectively
- In general, all \tilde{J}_j may be direction-dependent, so inside the integral....

Realizing practical calibration

- ...but in practice, we often ignore the direction dependence of the calibration components and factor them out of the integral (dropping \tilde{E}_{ij}). The Measurement Equation then becomes a relation between the observed and ideal visibilities:

$$\tilde{V}_{ij}^{obs} = \tilde{B}_{ij} \tilde{G}_{ij} \tilde{D}_{ij} \tilde{P}_{ij} \tilde{T}_{ij} \tilde{F}_{ij} \tilde{V}_{ij}^{ideal}$$

- If the ideal visibilities are known (e.g., by choosing calibration source of known structure), we can solve for individual components using those we already know (if any), e.g.:

$$(\tilde{G}_{ij}^{-1} \tilde{B}_{ij}^{-1} \tilde{V}_{ij}^{obs}) = \tilde{D}_{ij} (\tilde{P}_{ij} \tilde{T}_{ij} \tilde{F}_{ij} \tilde{V}_{ij}^{ideal})$$

Realizing practical calibration (cont)

- Formally, solving for any component is the same non-linear fitting problem:

$$(\tilde{V}_{ij}^{corrected-obs}) = \tilde{J}_{ij}^{solvable} (\tilde{V}_{ij}^{corrupted-ideal})$$

- Algebraic particulars are stored safely and conveniently inside the matrix formalism (out of sight, out of mind!)
- Viability of the solution relies on the underlying algebra (hard wired in calibration applications) and *proper calibration observations*
- The relative importance of the different components enables deferring or even ignoring the more subtle effects

Planning for Good Calibration

- A priori calibrations (provided by the observatory)
 - Antenna positions, earth orientation and rate
 - Clocks
 - Antenna pointing, gain, voltage pattern
 - Calibrator coordinates, flux densities, polarization properties
- Absolute calibration?
 - Very difficult, requires heroic efforts by visiting observers and observatory scientific and engineering staff
- Cross-calibration a better choice
 - Observe nearby point sources against which calibration components can be solved, and transfer solutions to target observations
 - Choose appropriate calibrators for different components; usually point sources because we can predict their visibilities
 - Choose appropriate timescales for each component

Calibrator Rules of Thumb

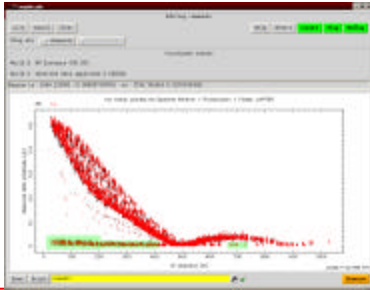
- T, G:**
 - Strong and point-like sources, as near to target source as possible
 - Observe often enough to track phase and amplitude variations: calibration intervals of up to 10s of minutes at low frequencies (beware of ionosphere!), as short as 1 minute or less at high frequencies
 - Observe at least one calibrator of known flux density at least once
- B:**
 - Strong enough for good sensitivity in each channel (often, T, G calibrator is ok)
 - If bandwidth is wide, should be point-like to avoid visibility changes across band
 - Observe often enough to track variations (e.g., waveguide reflections change with temperature and are thus a function of time-of-day)
- D:**
 - Best calibrator is strong and unpolarized
 - If polarized, observe over a broad range of parallactic angle to disentangle D_s and source polarization (often, T, G calibrator is ok)
- F:**
 - Choose strongly polarized source and observe often enough to track variation
 - If ionosphere is stable, rely on ionosphere observations for empirical corrections

Data Examination and Editing

- After observation, initial data examination and editing very important
 - Will observations meet goals for calibration and science requirements?
 - Some real-time flagging occurred during observation (antennas off-source, LO out-of-lock, etc.). Any such bad data left over? (check operator's logs)
 - Any persistently 'dead' antennas ($U_i=0$ during otherwise normal observing)? (check operator's logs)
 - Amplitude and phase should be continuously varying—edit outliers
 - Any antennas shadowing others? Edit such data.
 - Be conservative: those antennas/timeranges which are bad on calibrators are probably bad on weak target sources—edit them
 - Periods of poor weather? (check operator's log)
 - Distinguish between bad data and poorly-calibrated data. E.g., some antennas may have significantly different amplitude response which may not be fatal—it may only need to be calibrated
 - Radio Frequency Interference (RFI)?
 - Choose reference antenna wisely (ever-present, stable response)

A Data Editing Example

- msplot in aips++



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Radio Frequency Interference

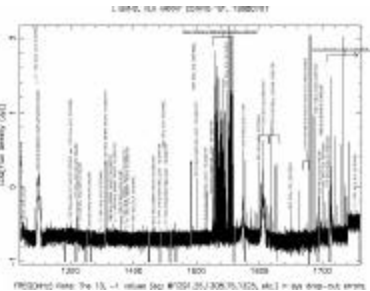
- RFI originates from man-made signals generated in the antenna or by external sources (e.g., satellites, cell-phones, radio and TV stations, etc.)
 - Obscures natural emission in spectral line observations
 - Adds to total noise power in all observations, thus decreasing sensitivity to desired natural signal, and complicating amplitude calibration
 - Though a contribution to the n_i term, can correlate between antennas if of common origin or baseline short enough
- RFI Mitigation
 - Careful electronics design in antennas
 - Observatories world-wide lobbying for spectrum management
 - Various on-line and off-line mitigation techniques under study
 - Choose interference-free frequencies (try to find 50 MHz of clean spectrum in the 1.6 GHz band!)
 - Observe continuum experiments in spectral-line modes so bad channels can be edited

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Radio Frequency Interference (cont.)

- Growth of telecom industry threatening radioastronomy!



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Calibration Sequence I

- Observation: total intensity spectral line imaging of weak target
 - A weak target source (1)
 - A nice near-by point-like G , T calibrator (2), observed alternately, but too weak for good B calibration (flux density unknown)
 - Three observations of strong flux density calibrator (3) which is also good for B calibration
- Schedule (each digit is a fixed duration):
33-2-1111-2-1111-2-1112-1112-2-1112-1112-1112-1112-2-33
- Calibration sequence:
 - On 3, solve for G : $\langle \tilde{V}_{(0,1)}^{obs} \rangle = \tilde{G}_{(0,1)} \langle \tilde{V}_{(0,1)}^{ideal} \rangle$
 - On 3, solve for B , using G : $\langle \tilde{V}_{(0,1)}^{obs} \rangle = \tilde{B}_{(0,1)} \langle \tilde{G}_{(0,1)} \tilde{V}_{(0,1)}^{ideal} \rangle$
 - On 2, solve for G , using B : $\langle \tilde{R}_{(0,1)}^{obs} \rangle = \tilde{G}_{(0,1)} \langle \tilde{V}_{(0,1)}^{ideal} \rangle$
 - Scale 2's G s according to 3's G s: $\langle \tilde{G}_{(0,1)}^{obs} \rangle = \langle \tilde{G}_{(0,1)} \rangle \langle \tilde{G}_{(0,1)}^{obs} \rangle / \langle \tilde{G}_{(0,1)} \rangle$
 - Transfer B , G to 1: $\langle \tilde{V}_{(0,1)}^{obs} \rangle = \langle \tilde{G}_{(0,1)}^{obs} \rangle \langle \tilde{B}_{(0,1)} \rangle \langle \tilde{V}_{(0,1)}^{ideal} \rangle$

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Calibration Sequence II

- Observation: full-polarization imaging of weak target
 - A weak target source (1)
 - A nice, near-by, point-like, G , T calibrator (2), observed alternately, ok for D calibration, too (flux density and polarization unknown)
 - Three observations of strong flux density calibrator
- Schedule (each digit is a fixed duration):
33-2-1111-2-1111-2-1112-1112-2-1112-1112-1112-1112-2-33
- Calibration sequence:
 - On 2 & 3, solve for G , using P : $\langle \tilde{V}_{(0,1)}^{obs} \rangle = \tilde{G}_{(0,1)} \langle \tilde{P}_{(0,1)} \rangle \langle \tilde{V}_{(0,1)}^{ideal} \rangle$
 - Apply G to 2, get improved poln model: $\langle \tilde{P}_{(0,1)}^{ideal} \rangle = \langle \tilde{P}_{(0,1)} \rangle \langle \tilde{V}_{(0,1)}^{obs} \rangle / \langle \tilde{V}_{(0,1)}^{ideal} \rangle$
 - On 2, solve for D , using P , G , and new model: $\langle \tilde{D}_{(0,1)}^{obs} \rangle = \langle \tilde{D}_{(0,1)} \rangle \langle \tilde{P}_{(0,1)} \rangle \langle \tilde{V}_{(0,1)}^{ideal} \rangle$
 - Scale 2's G s according to 3's G s: $\langle \tilde{G}_{(0,1)}^{obs} \rangle = \langle \tilde{G}_{(0,1)} \rangle \langle \tilde{D}_{(0,1)}^{obs} \rangle / \langle \tilde{D}_{(0,1)} \rangle$
 - Transfer D , G to 1, use P : $\langle \tilde{V}_{(0,1)}^{obs} \rangle = \langle \tilde{G}_{(0,1)}^{obs} \rangle \langle \tilde{D}_{(0,1)} \rangle \langle \tilde{P}_{(0,1)} \rangle \langle \tilde{V}_{(0,1)}^{ideal} \rangle$

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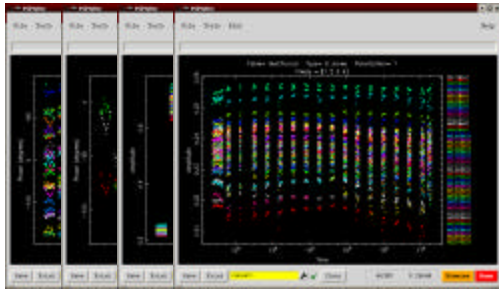
Evaluating Calibration Performance

- Are solutions continuous?
 - Noise-like solutions are just that—noise
 - Discontinuities indicate instrumental glitches
 - Any additional editing required?
- Are calibrator data fully described by antenna-based effects?
 - Phase and amplitude *closure errors* are the baseline-based residuals; see Chapter 5 in book
 - Are calibrators sufficiently point-like? If not, self-calibrate: model calibrator visibilities (by imaging, deconvolving and transforming) and re-solve for calibration; iterate to isolate source structure from calibration components
 - » Jim's lecture: "Self-Calibration" (Wednesday)
- Any evidence of unsampled variation? Is interpolation of solutions appropriate?
 - Self-calibration may be required, if possible

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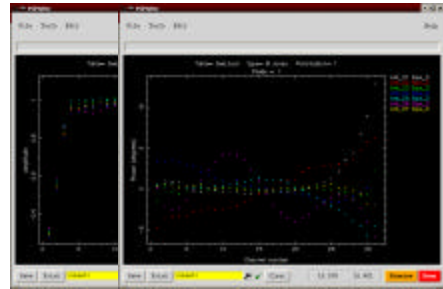
G Solution Examples



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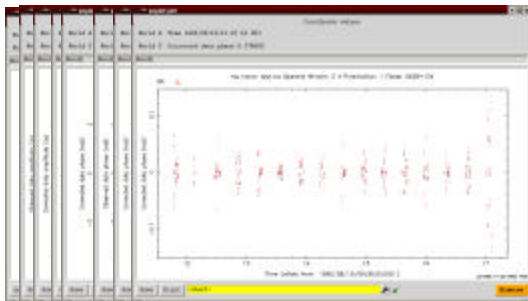
B Solution Examples



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Effect of Calibration on Visibility Data

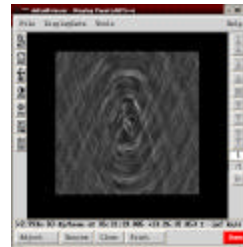


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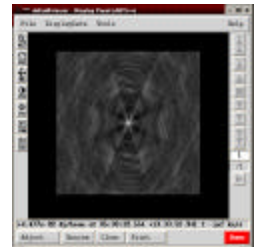
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Effect of Calibration in the Image Plane

Uncalibrated



Calibrated



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Summary

- Determining calibration is as important as determining source structure—can't have one without the other
- Data examination and editing and important part of calibration
- Calibration dominated by antenna-based effects
- Calibration formalism algebra-rich, but can be described piecemeal in comprehensible segments, according to well-defined effects
- Calibration determination is a single standard fitting problem
- Point sources are the best calibrators
- Observe calibrators according requirements of components
- Calibration sequences a juggling act of effects and corrections

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